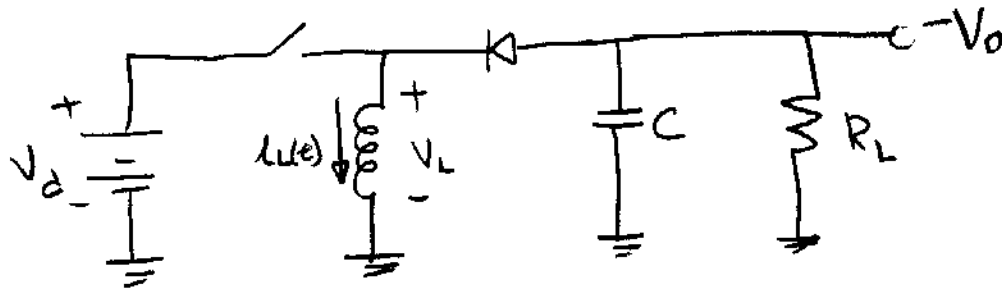
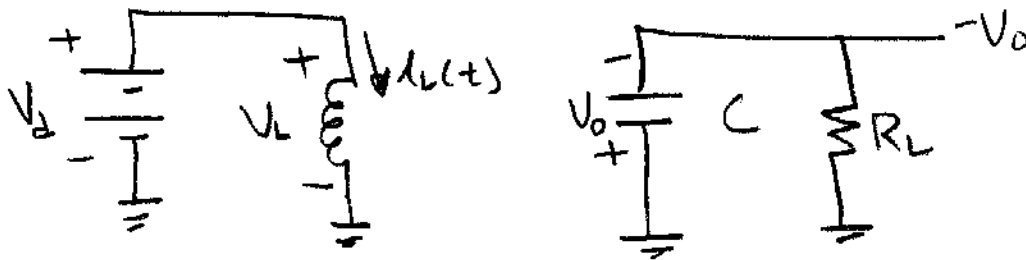


# Polarity Inverting Regulator OR Buck-Boost Regulator



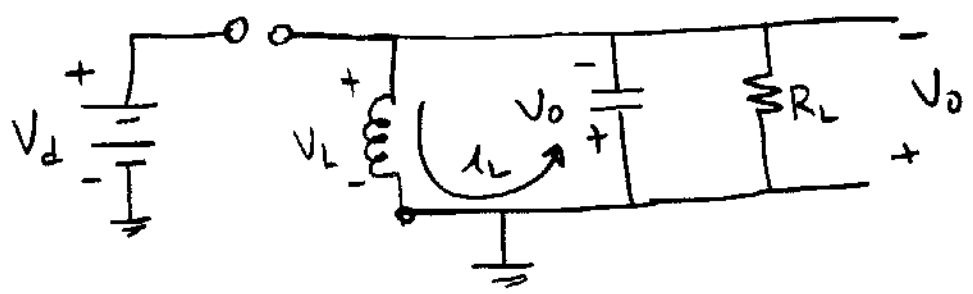
During  $t_{on}$ : Switch Closed  
D = off



- Capacitor supplies power to load

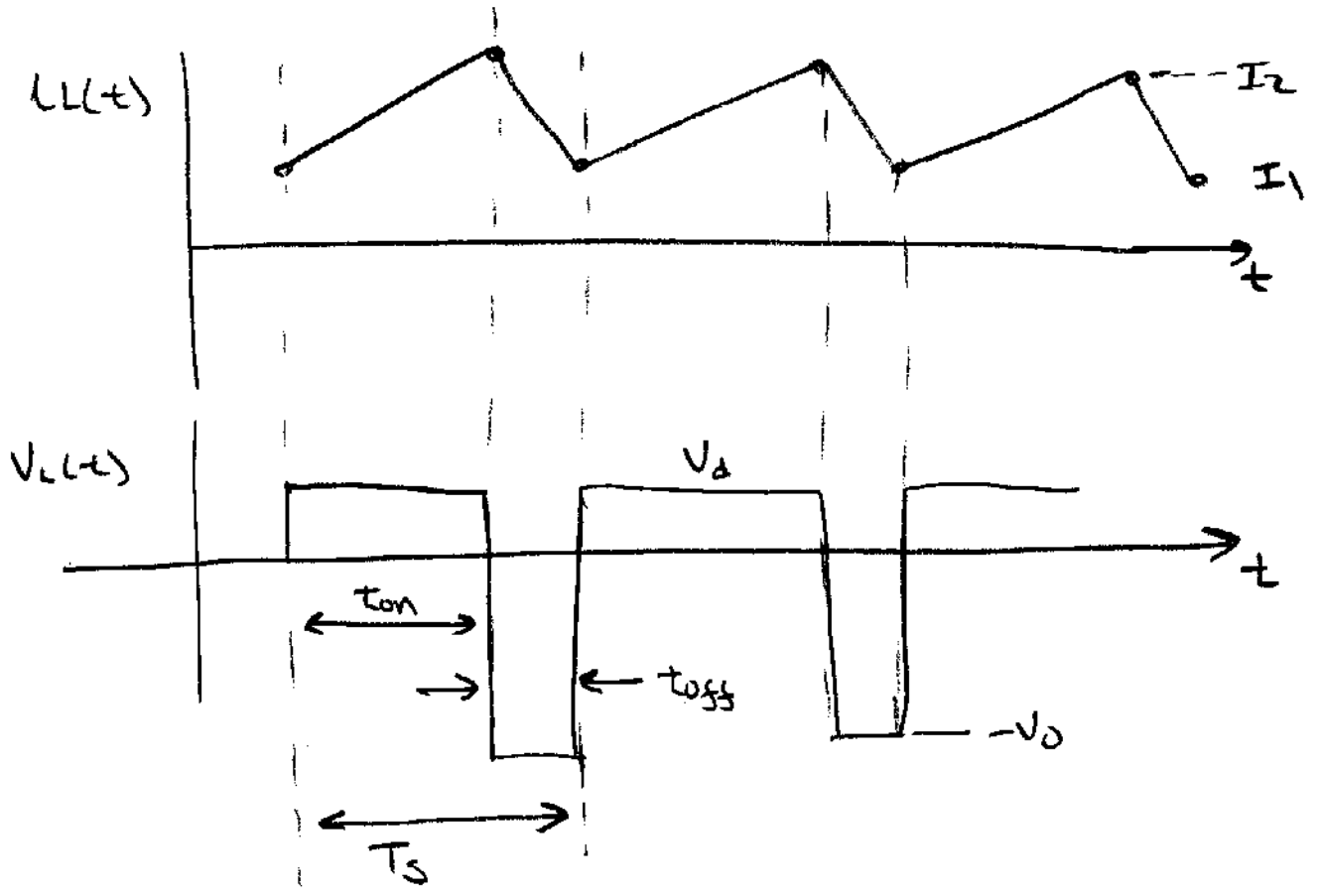
$$V_L = V_d$$

During  $t_{off}$  : Switch = off  
 $D = ON$



- $V_L = -V_o$
- Capacitor supplies power to Load
- Inductor dumps energy to capacitor

CONTINUOUS mode operation



- In Continuous mode  $I_L > 0$

and  $t_{on} + t_{off} = T_s$

- Since in steady state, the average current over one cycle is constant,

$$\int_{T_s} V_L(t) dt = 0$$

OR  $V_d t_{on} + (-V_o t_{off}) = 0$

OR  $V_d t_{on} - V_o (T_s - t_{on}) = 0$

OR  $\boxed{\frac{V_o}{V_d} = \frac{t_{on}}{T_s - t_{on}} = \frac{D}{1-D}}$  ;  $D = \frac{t_{on}}{T_s}$  (1)

Assuming a lossless circuit,  $P_d = P_o$ , OR

$$V_d I_d = V_o I_o$$

Where  $I_d$  = average input current and

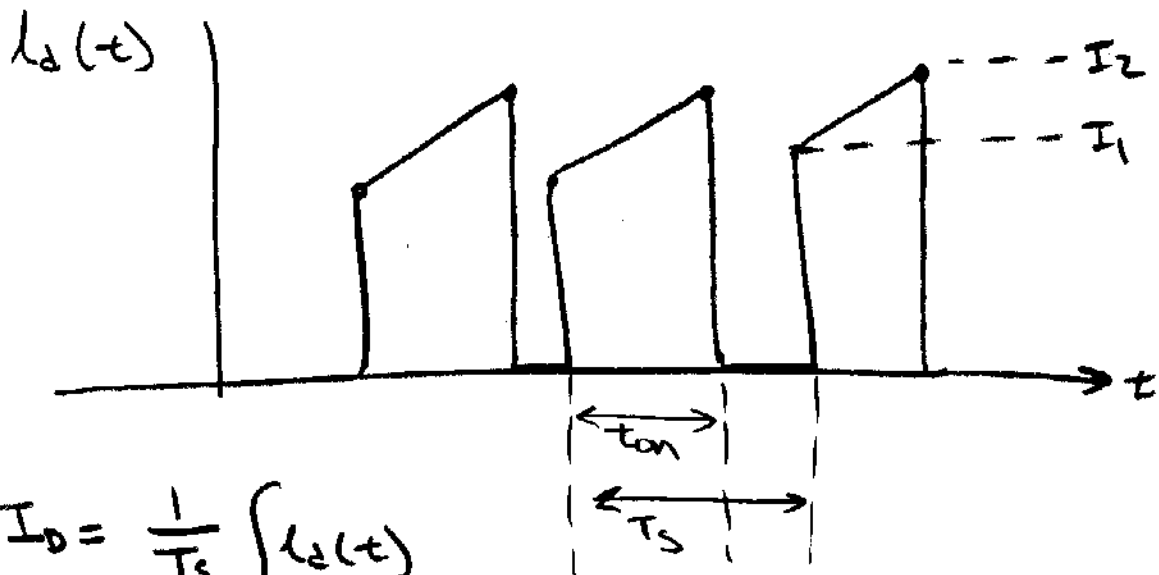
$I_o$  = average output current

Since  $P_d = P_o$

$$\frac{I_o}{I_d} = \frac{V_d}{V_o} = \frac{1-D}{D} \quad (2)$$

- Now find the average input current

-  $V_d$  only supplies current during  $t_{on}$



$$I_o = \frac{1}{T_s} \int I_d(t)$$

$$I_o = \left( \frac{I_1 + I_2}{2} \right) \frac{t_{on}}{T_s} = \left( \frac{I_1 + I_2}{2} \right) D$$

Sub into eq (2)

$$I_0 = \frac{1-D}{D} I_2$$

$$I_0 = \frac{1-D}{D} \left( \frac{I_1 + I_2}{2} \right) D$$

$$I_1 + I_2 = \frac{2 I_0}{1-D}$$

③

Next use  $V_L(t) = L \frac{di_L}{dt}$

$$i_L(t) = \frac{1}{L} \int V_L dt + i.c.$$

$$I_2 = \frac{1}{L} \int_0^{t_{on}} V_d dt + I_1$$

$$I_2 - I_1 = \frac{V_d t_{on}}{L}$$

④

- Next, Find the boundary between continuous + discontinuous mode.

=> solve for  $I_1$

$$I_1 + I_2 = \frac{2 I_0}{1-D}$$

(-)

$$I_2 - I_1 = \frac{V_d t_{on}}{L}$$

---

$$2 I_1 = \left( \frac{2 I_0}{1-D} \right) - \frac{V_d t_{on}}{L}$$

For the boundary btw continuous and discontinuous mode, solve for  $I_1 = 0$

$$0 = \frac{2 I_0}{1-D} - \frac{V_d t_{on}}{L}$$

$$I_0 = \left( \frac{V_d t_{on}}{2L} \right) (1-D)$$

OR

$$I_0 \geq \frac{V_d t_{on}}{2L} (1-D)$$

## Summary

BUCK-Boost in  
continuous mode

$$V_o = V_d \left[ \frac{D}{1-D} \right]$$

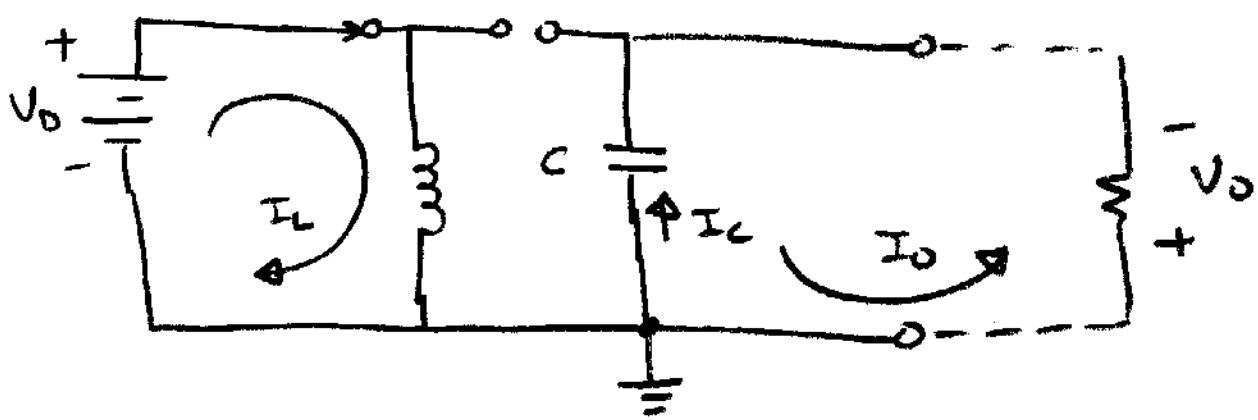
$$I_o = I_b \left[ \frac{1-D}{D} \right]$$

$$I_1 + I_2 = \frac{2 I_o}{1-D}$$

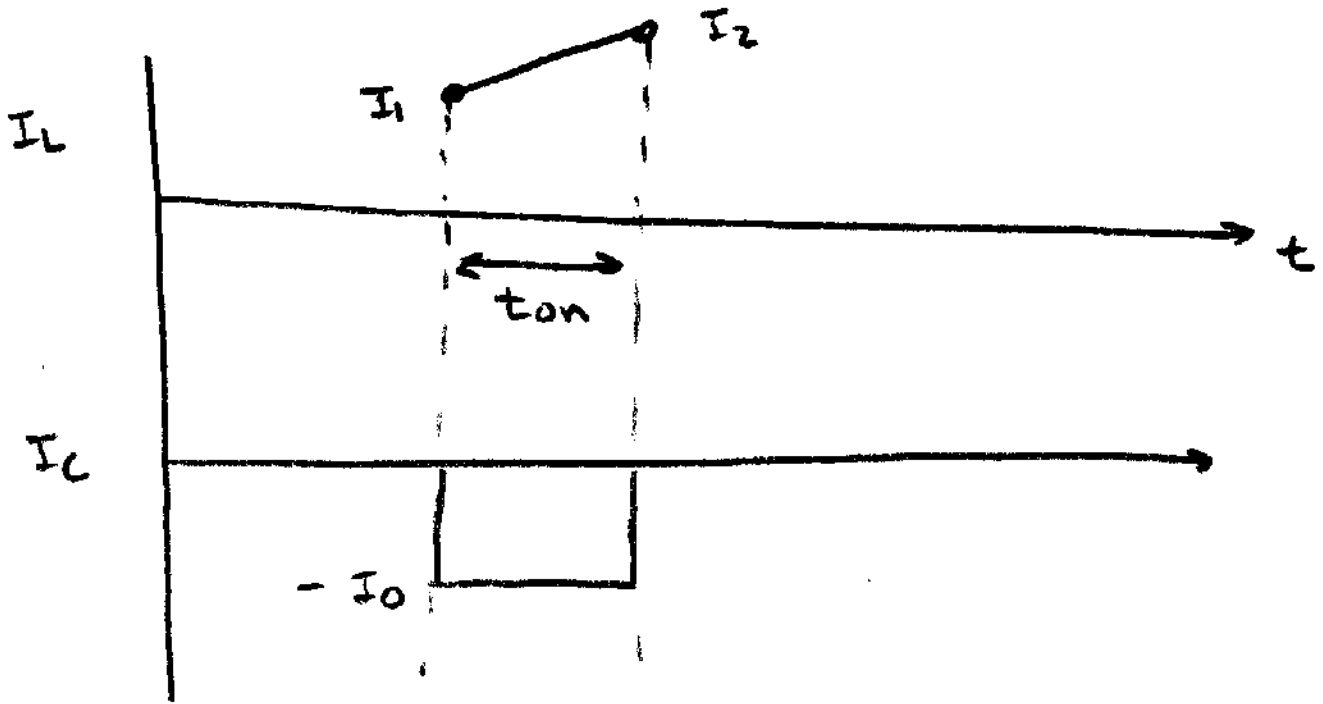
$$I_2 - I_1 = \frac{V_d t_{on}}{L}$$

$$I_o \geq \frac{V_d t_{on}}{2L} (1-D)$$

- Calculation of Capacitor Ripple Current
- During  $t_{on}$ , the diode is off and the capacitor must supply the entire output current

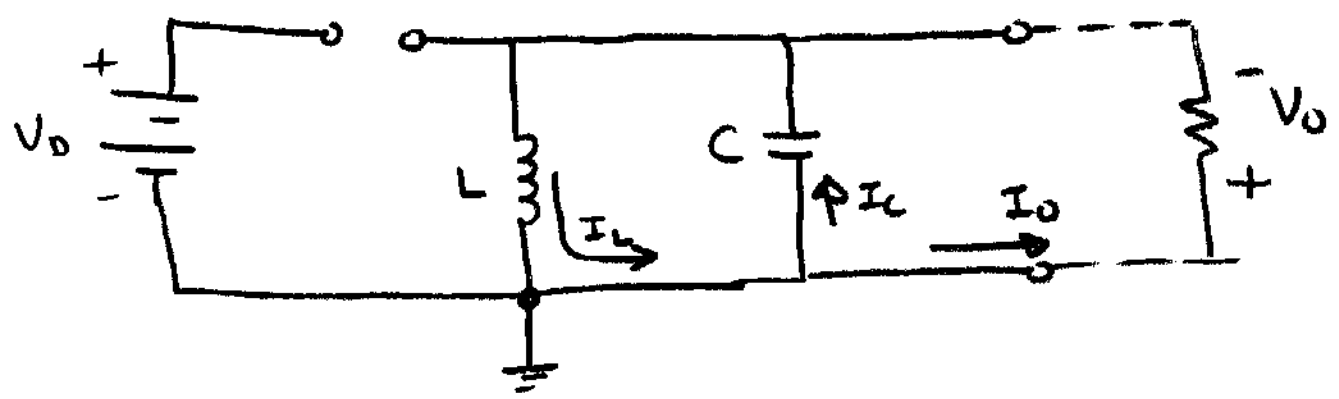


During  $t_{on}$  The inductor charges from  $I_1$  to  $I_2$

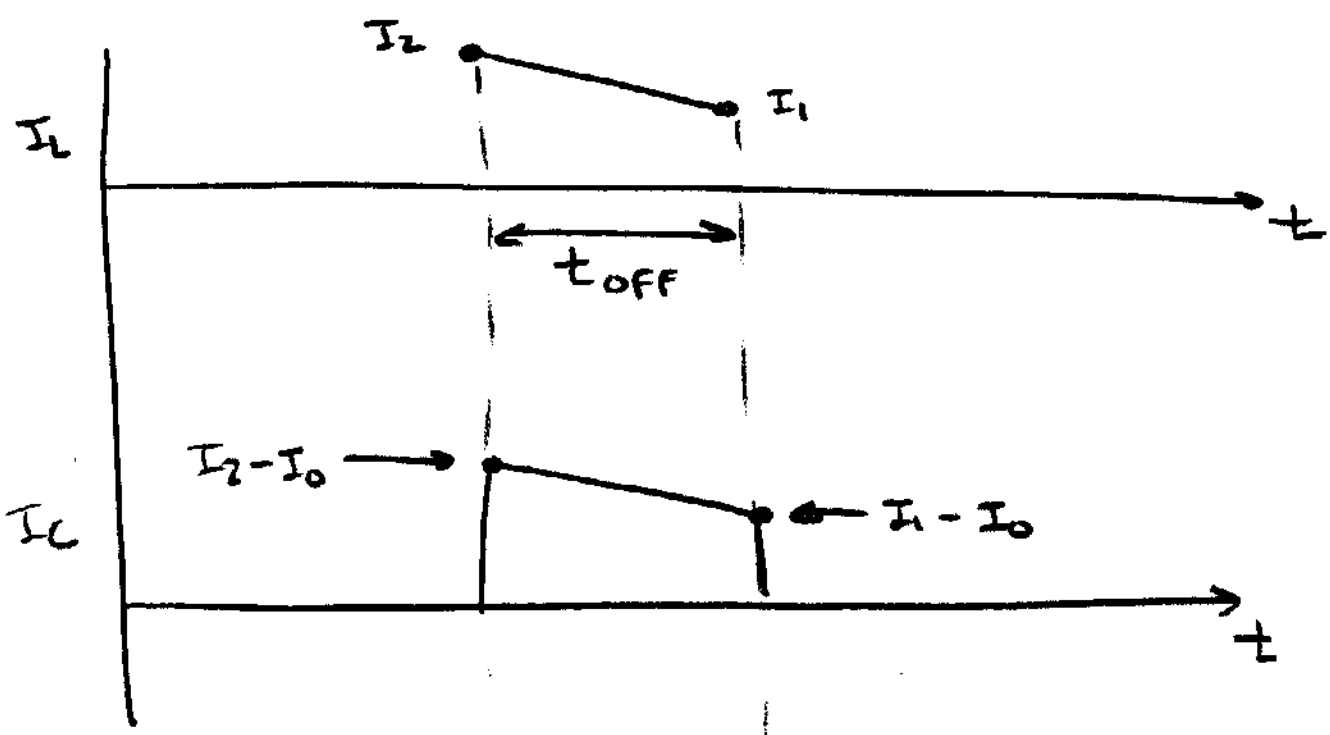


- During  $t_{off}$ , the inductor charges the capacitor and supplies current to the output

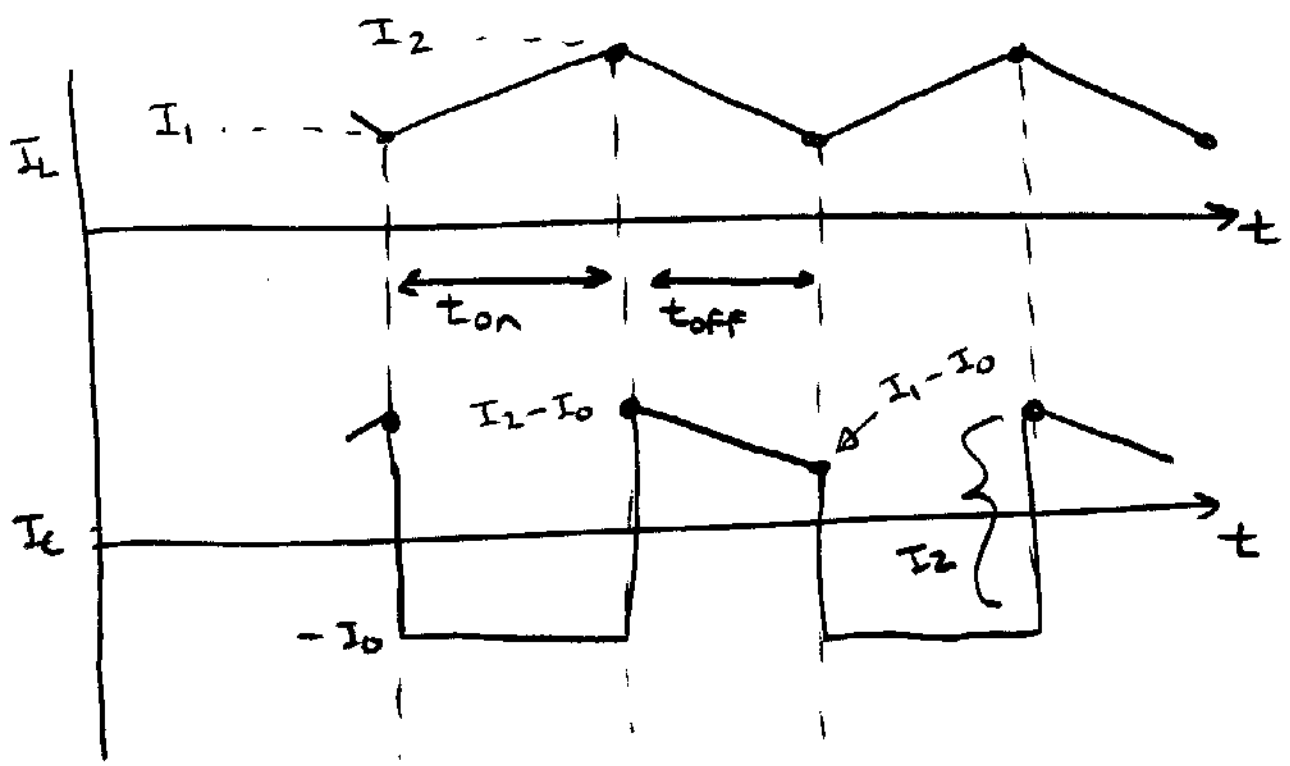
$$\Rightarrow I_C = I_L - I_O$$



- During  $t_{off}$ , the inductor discharges from  $I_2$  to  $I_1$



- The complete picture is



$$I_{Rms} = \sqrt{\frac{1}{T_s} \left[ \int_0^{t_{on}} (-I_0)^2 dt + \int_0^{t_{off}} (I_L - I_0) dt \right]}$$

during  $t_{off}$ , we can express  $I_L(t)$  as

$$I_L(t) = I_2 + m t$$

where  $m = \frac{I_1 - I_2}{t_{off}}$

Also: Ripple due to ESR is

$$V_{CR} = I_2 \cdot ESR$$

# EE 456

## Buck-Boost Regulator Design Continuous Mode Operation

$$\mu\text{s} := 10^{-6} \cdot \text{sec} \quad \text{m}\Omega := \frac{\Omega}{1000}$$

Specify Input Voltage

Specify Output Voltage

$$V_D := 12 \cdot \text{volt}$$

$$V_o := -12 \cdot \text{volt}$$

Specify Switching Frequency

$$F_S := 40 \cdot \text{kHz}$$

$$T_S := \frac{1}{F_S} \quad T_S = 25 \mu\text{s}$$

Specify the Assumed Efficiency

$$\text{Eff} := 85\%$$

Specify the Max output Current

The output Power is

$$P_{\text{out}} := \frac{50 \cdot \text{watt}}{\text{Eff}}$$

The output current is

$$I_o := \left| \frac{P_{\text{out}}}{V_o} \right| \quad I_o = 4.902 \text{ amp}$$

Find Ton

$$t_{\text{on}} := 1 \cdot \mu\text{s}$$

Given

$$-V_o = V_D \cdot \left( \frac{t_{\text{on}}}{T_S - t_{\text{on}}} \right)$$

$$t_{\text{on}} := \text{Find}(t_{\text{on}})$$

$$t_{\text{on}} = 12.5 \mu\text{s}$$

$$t_{\text{off}} := T_S - t_{\text{on}}$$

$$t_{\text{off}} = 12.5 \mu\text{s}$$

$$D := \frac{t_{\text{on}}}{T_S}$$

$$D = 50\%$$

Find the range of Inductors that will operate in continuous mode

Specify the minimum current we want the supply to operate in the continuous mode

$$I_{\text{min}} := \frac{I_o}{10}$$

$$L := \frac{V_D \cdot t_{\text{on}}}{2 \cdot I_{\text{min}}} \cdot (1 - D)$$

For Continuous Mode, We need L greater than

$$L = 76.5 \mu\text{H}$$

Choose the Inductor  $L := 90 \cdot \mu\text{H}$

Pulse  
Engineering  
PE-51512

Find the Min and max inductor currents

$$I_1 := 1 \cdot \text{amp} \quad I_2 := 1 \cdot \text{amp}$$

Given

$$\frac{I_1 + I_2}{2} = \frac{I_o}{1 - D}$$

$$I_2 - I_1 = V_D \cdot \frac{t_{\text{on}}}{L}$$

$$\begin{pmatrix} I_1 \\ I_2 \end{pmatrix} := \text{Find}(I_1, I_2) \quad I_1 = 8.971 \text{ amp} \quad I_2 = 10.637 \text{ amp}$$

Inductor Average Current is  $\frac{I_1 + I_2}{2} = 9.804 \text{ amp}$

Choose the filter capacitor using the capacitor ESR. Assume that the major component of the ripple comes from the capacitor ESR.

Specify the ripple due to the ESR  $V_{CR} := 100 \cdot \text{mV}$

$$\text{ESR} := \frac{V_{CR}}{I_2} \quad \text{ESR} = 9.401 \text{ m}\Omega$$

For all electrolytic caps, assume that  $\text{ESR} \cdot C = 80 \text{ms}$

$$C := \frac{80 \cdot \mu\text{s}}{\text{ESR}} \quad C = 8510 \mu\text{F}$$

Choose the next size std capacitor  $C := 10000 \cdot \mu\text{F}$

Calculate the Capacitor RMS Ripple Current

Define a function for the inductor current  $I_L(t) := I_2 + \frac{(I_1 - I_2)}{t_{\text{off}}} \cdot t$

$$I_{\text{rms}} := \sqrt{\frac{1}{T_S} \cdot \left[ \int_{0 \cdot \text{sec}}^{t_{\text{on}}} (-I_0)^2 dt + \int_{0 \cdot \text{sec}}^{t_{\text{off}}} (I_L(t) - I_0)^2 dt \right]}$$

$$I_{\text{rms}} = 4.914 \text{ amp}$$

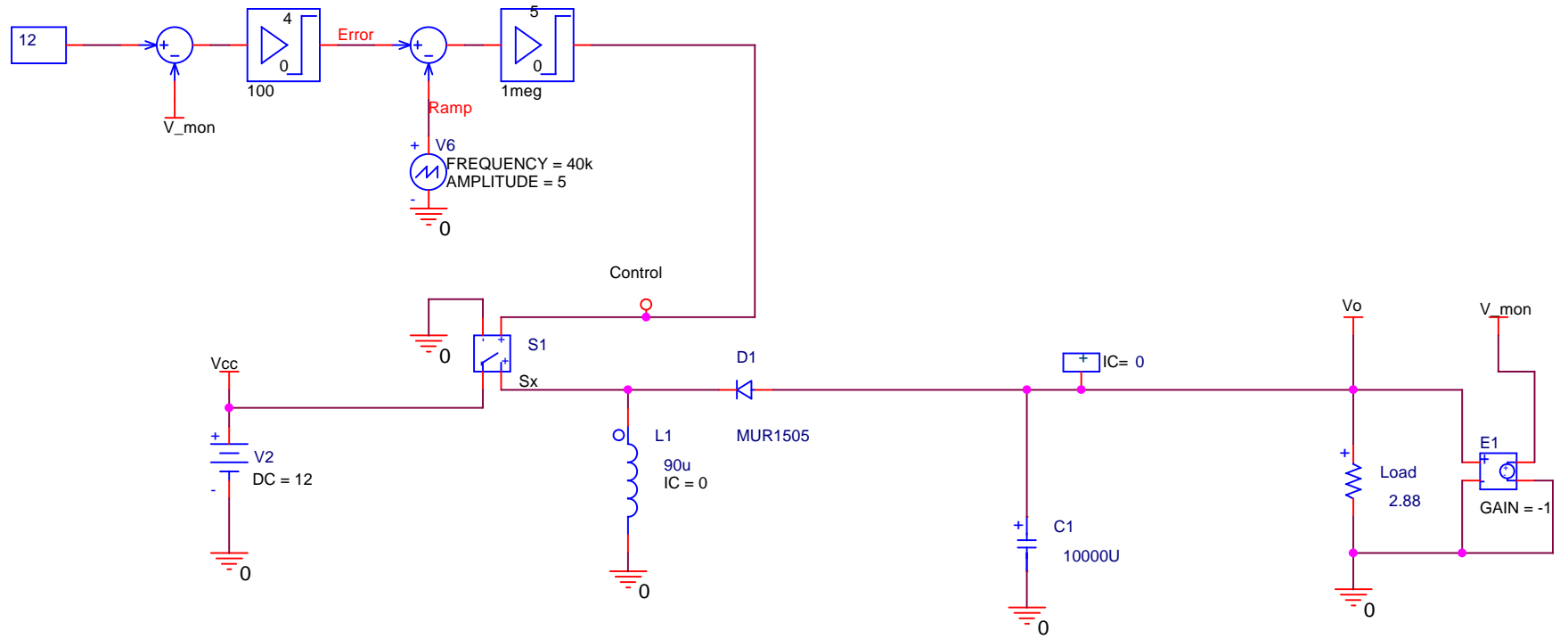
## Summary

$$L = 90 \mu\text{H} \quad I_2 = 10.637 \text{ amp} \quad \frac{I_2 + I_2}{2} = 10.637 \text{ amp}$$


$$t_{\text{on}} = 12.5 \mu\text{s} \quad t_{\text{off}} = 12.5 \mu\text{s}$$

$$V_D = 12 \text{ volt} \quad V_o = -12 \text{ volt} \quad I_o = 4.902 \text{ amp}$$

$$C = 10000 \mu\text{F} \quad V_{\text{CR}} = 100 \text{ mV} \quad I_{\text{rms}} = 4.914 \text{ amp}$$



Model = Sx VSWITCH Roff=1e6 Ron=1m Voff=1 Von=4.0

		ECE Department 5500 Wabash Avenue Terre Haute, IN 47803 Ph: (812) 877-8512 FAX: (253) 369-9536	
		Name: Marc E. Herniter      Class: ECE456	
Size A	Document Name		Rev 1
Date: Thursday, January 09, 2003		Sheet 1 of 1	

### Simulation Settings - TRans

General | Analysis | Include Files | Libraries | Stimulus | Options | Data Collection | Probe Window

Analysis type: **Time Domain (Transient)**

Run to time:  seconds (TSTOP)

Start saving data after:  seconds

Options:

- General Settings
- Monte Carlo/Worst Case
- Parametric Sweep
- Temperature (Sweep)
- Save Bias Point
- Load Bias Point

Transient options:

Maximum step size:  seconds

Skip the initial transient bias point calculation (SKIPBP)

[Output File Options...](#)

OK Cancel Apply Help

### Simulation Settings - TRans

General | Analysis | Include Files | Libraries | Stimulus | Options | Data Collection | Probe Window

Category: **Analog Simulation**

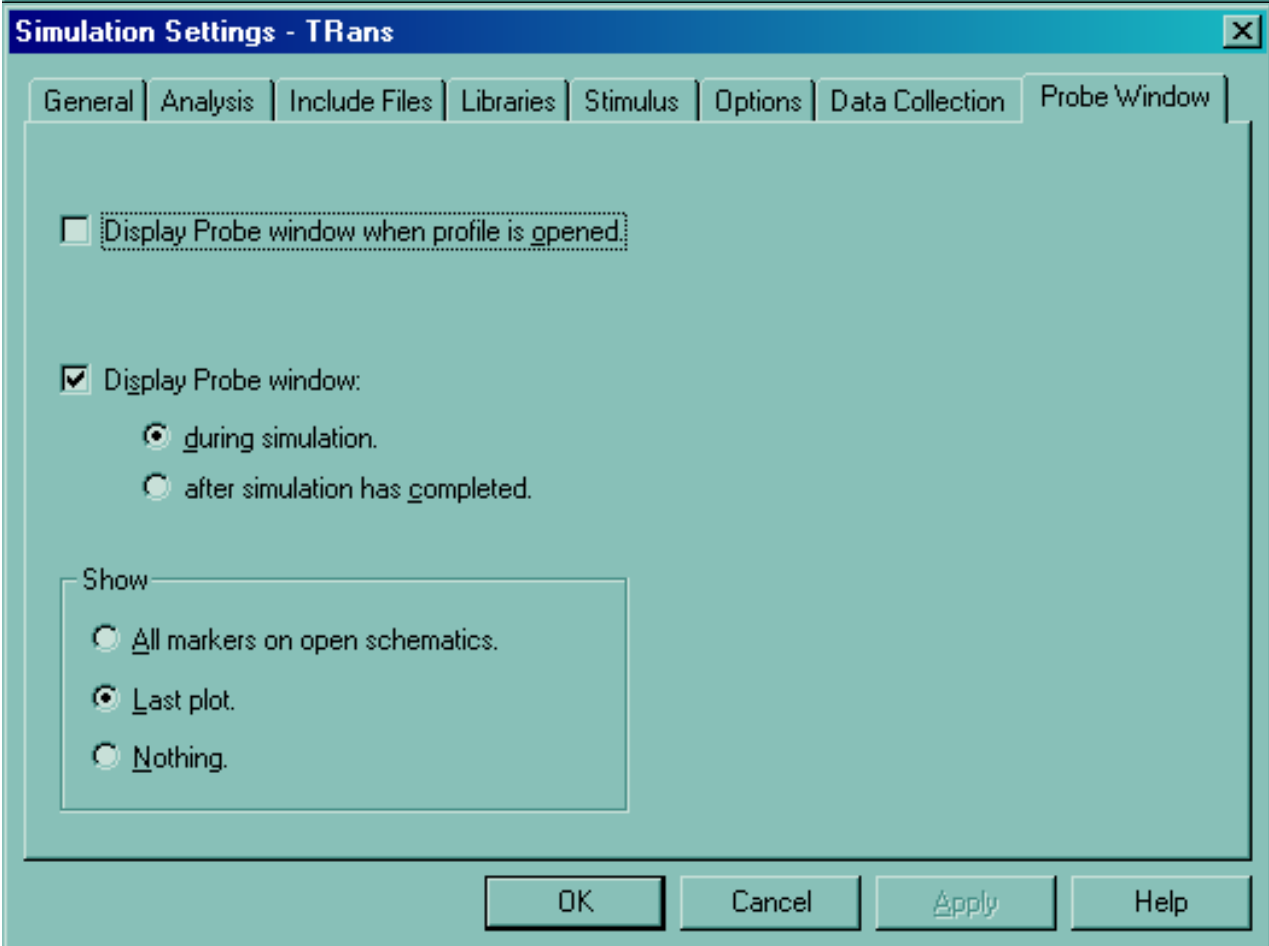
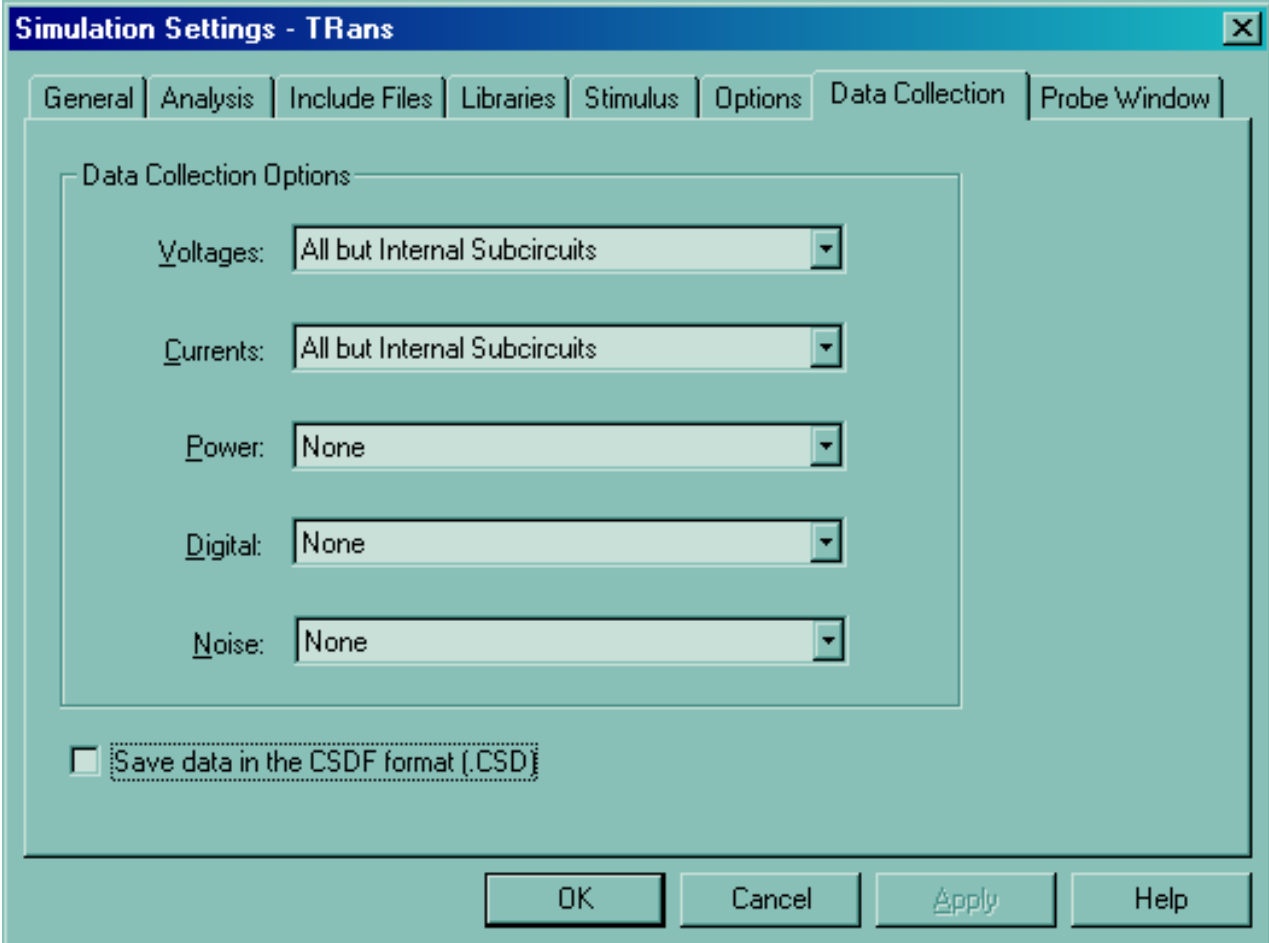
Relative accuracy of V's and I's:	<input type="text" value="0.001"/>		(.OPTION)
Best accuracy of voltages:	<input type="text" value="1.0u"/>	volts	(VNTOL)
Best accuracy of currents:	<input type="text" value="1.0p"/>	amps	(ABSTOL)
Best accuracy of charges:	<input type="text" value="0.01p"/>	coulombs	(CHGTOL)
Minimum conductance for any branch:	<input type="text" value="1.0E-12"/>	1/ohm	(GMIN)
DC and bias "blind" iteration limit:	<input type="text" value="150"/>		(ITL1)
DC and bias "best guess" iteration limit:	<input type="text" value="20"/>		(ITL2)
Transient time point iteration limit:	<input type="text" value="200"/>		(ITL4)
Default nominal temperature:	<input type="text" value="27.0"/>	°C	(TNOM)

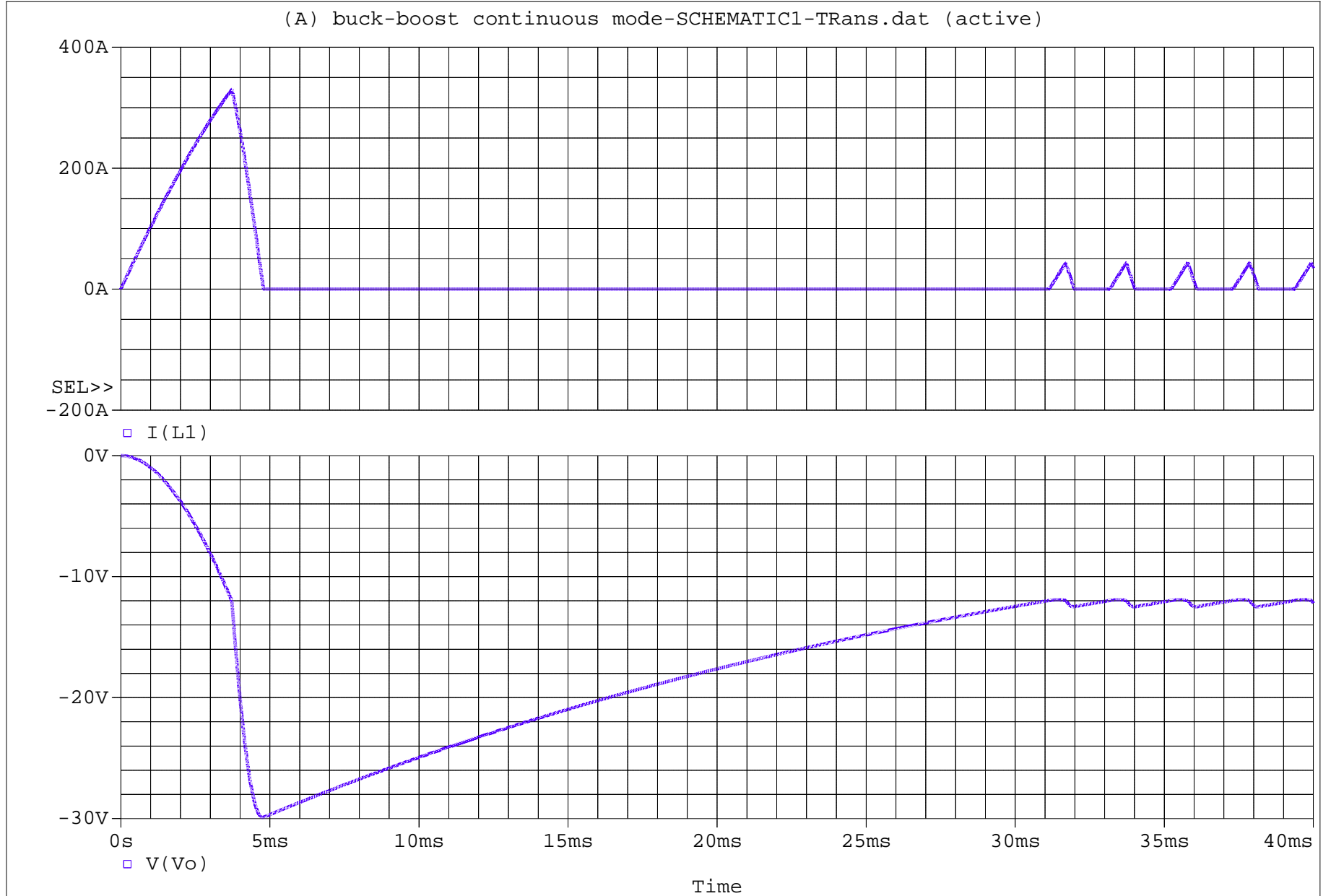
Use GMIN stepping to improve convergence. (STEPGMIN)

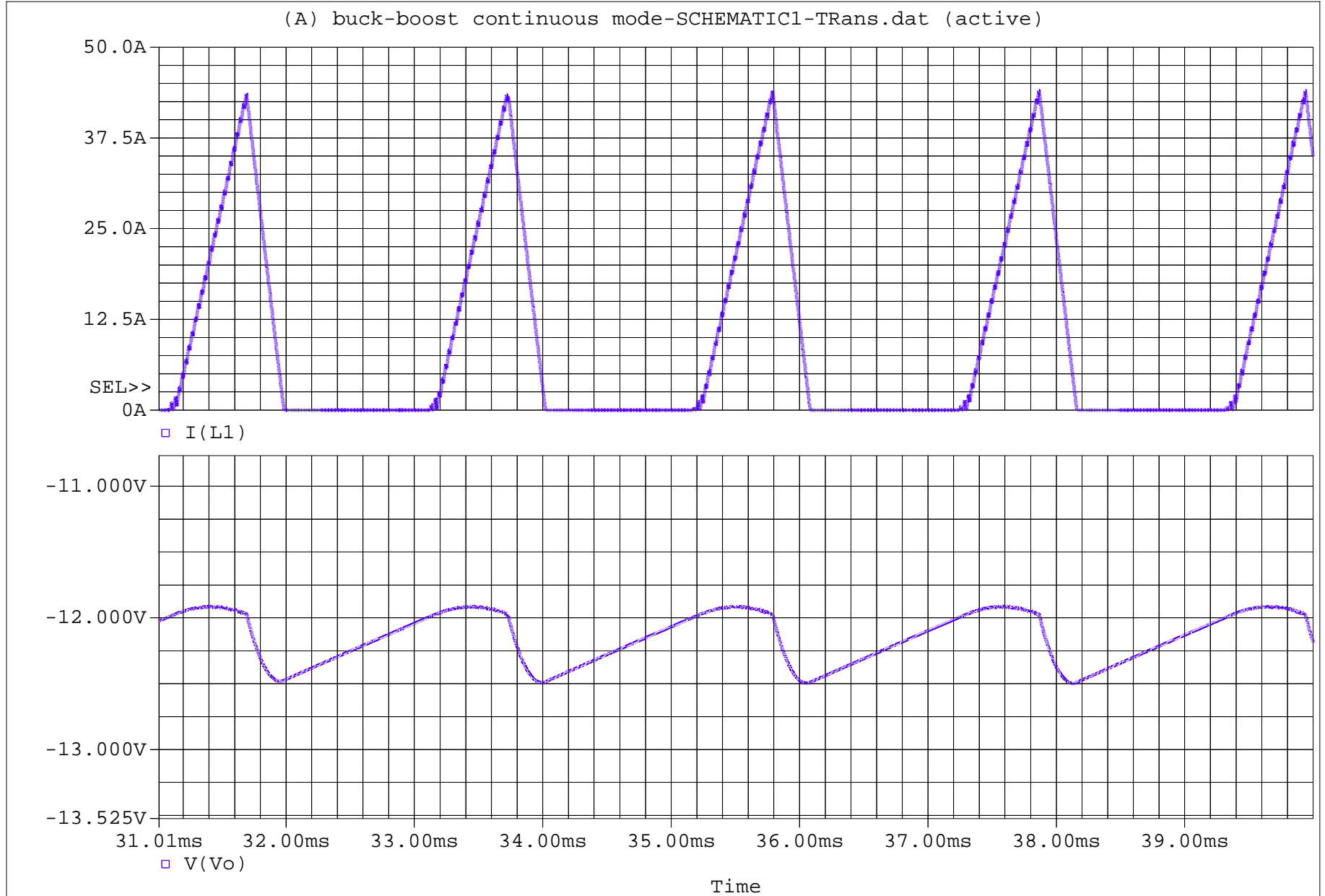
Use reordering to reduce matrix fill-in. (PREORDER)

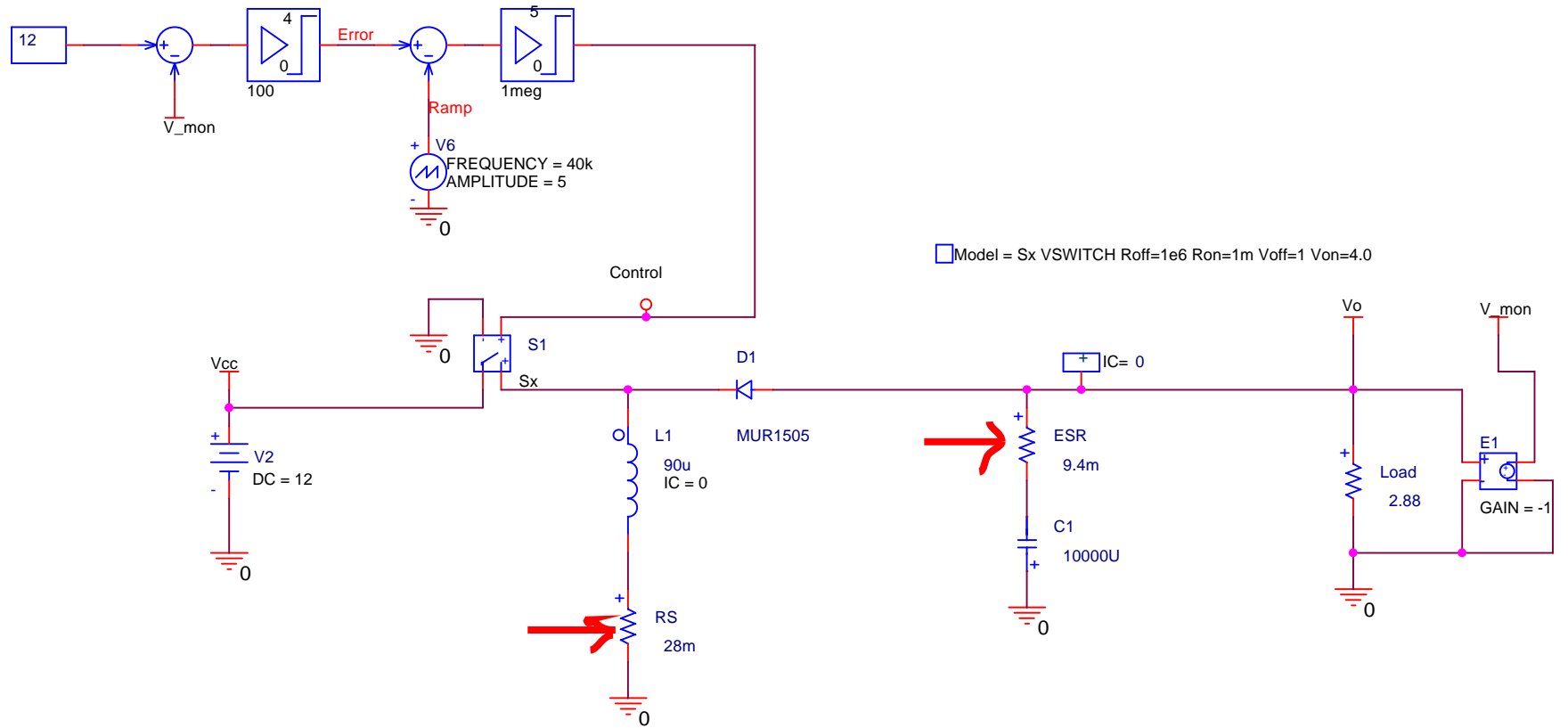
[MOSFET Options...](#) [Advanced Options...](#) [Reset](#)

OK Cancel Apply Help





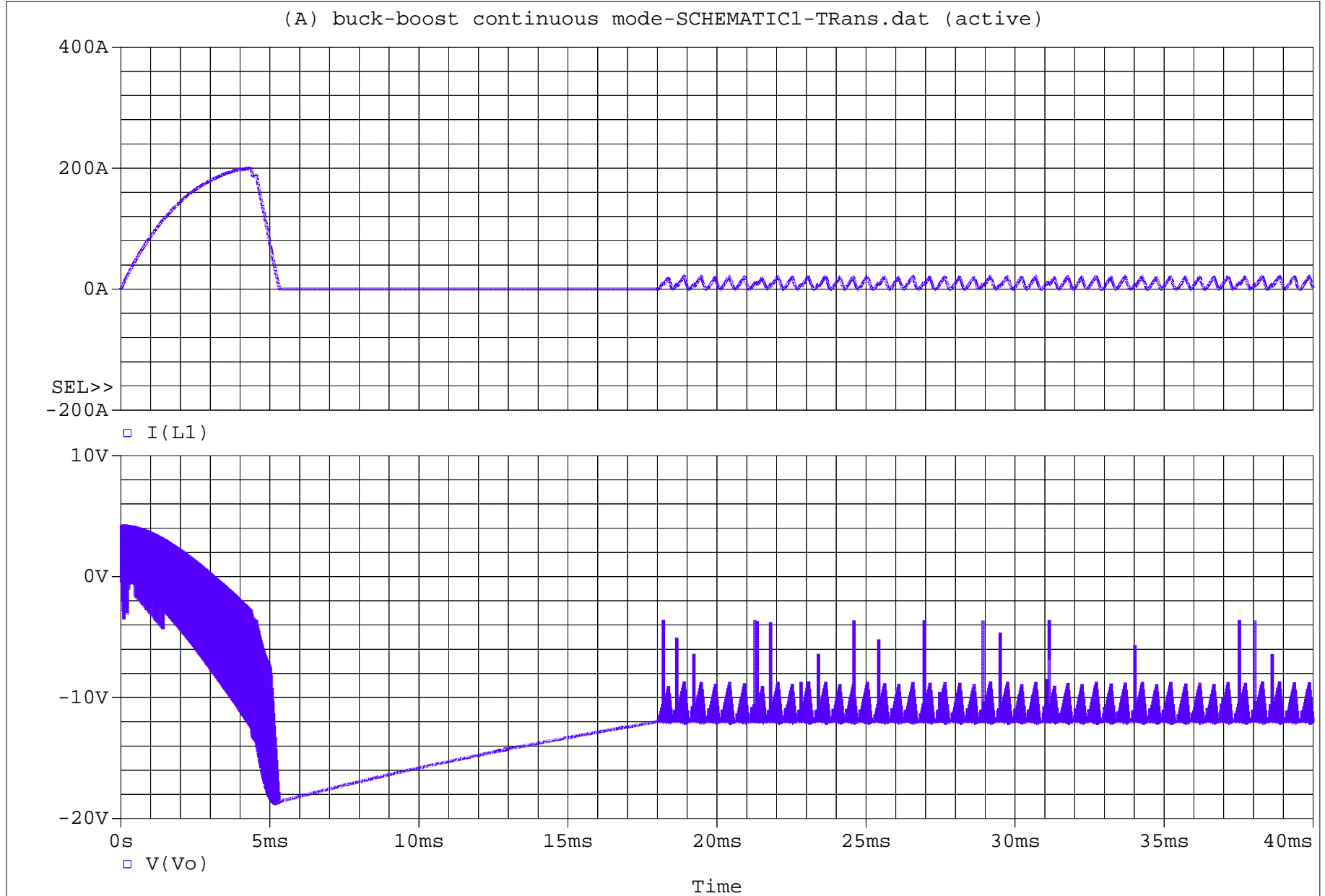


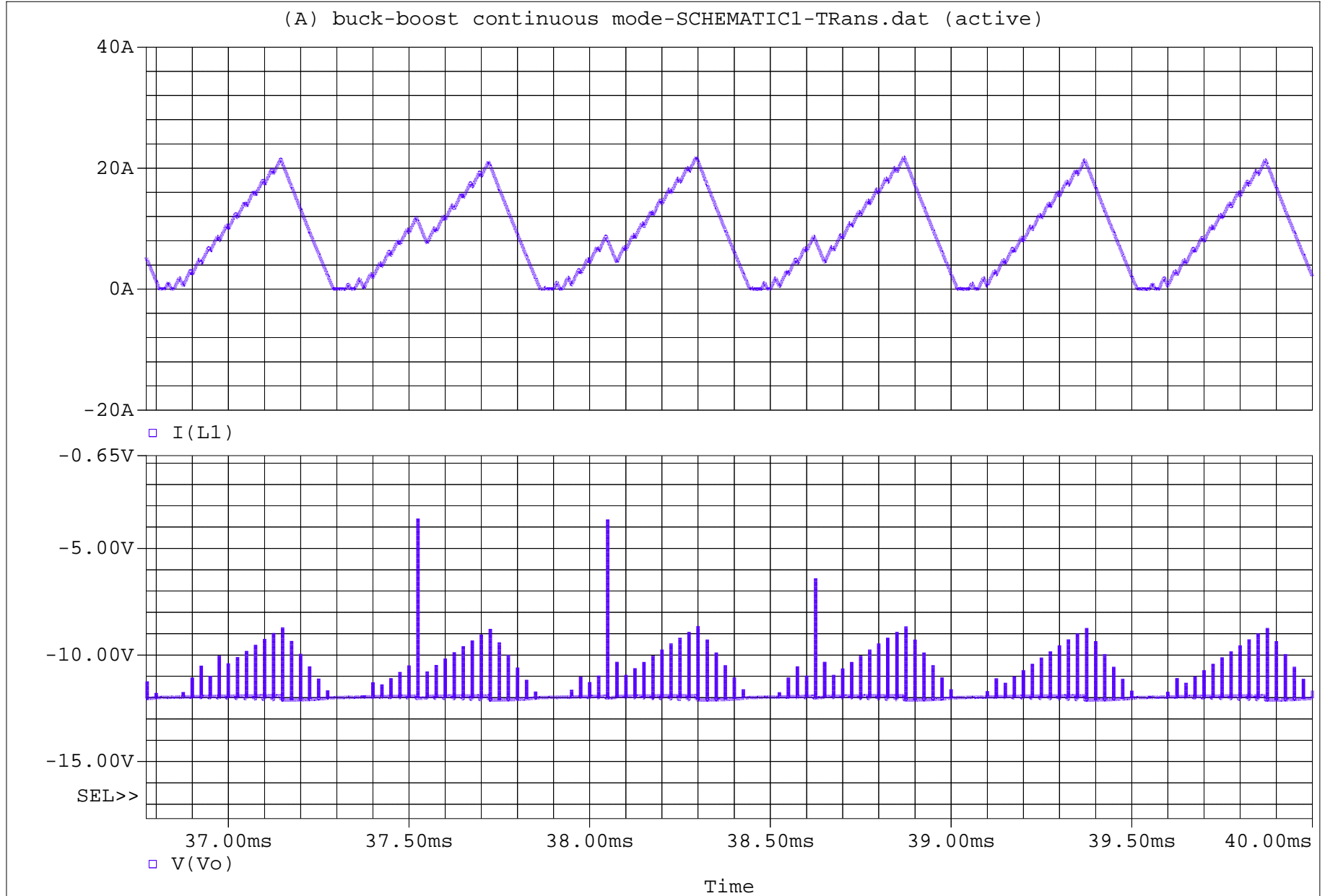


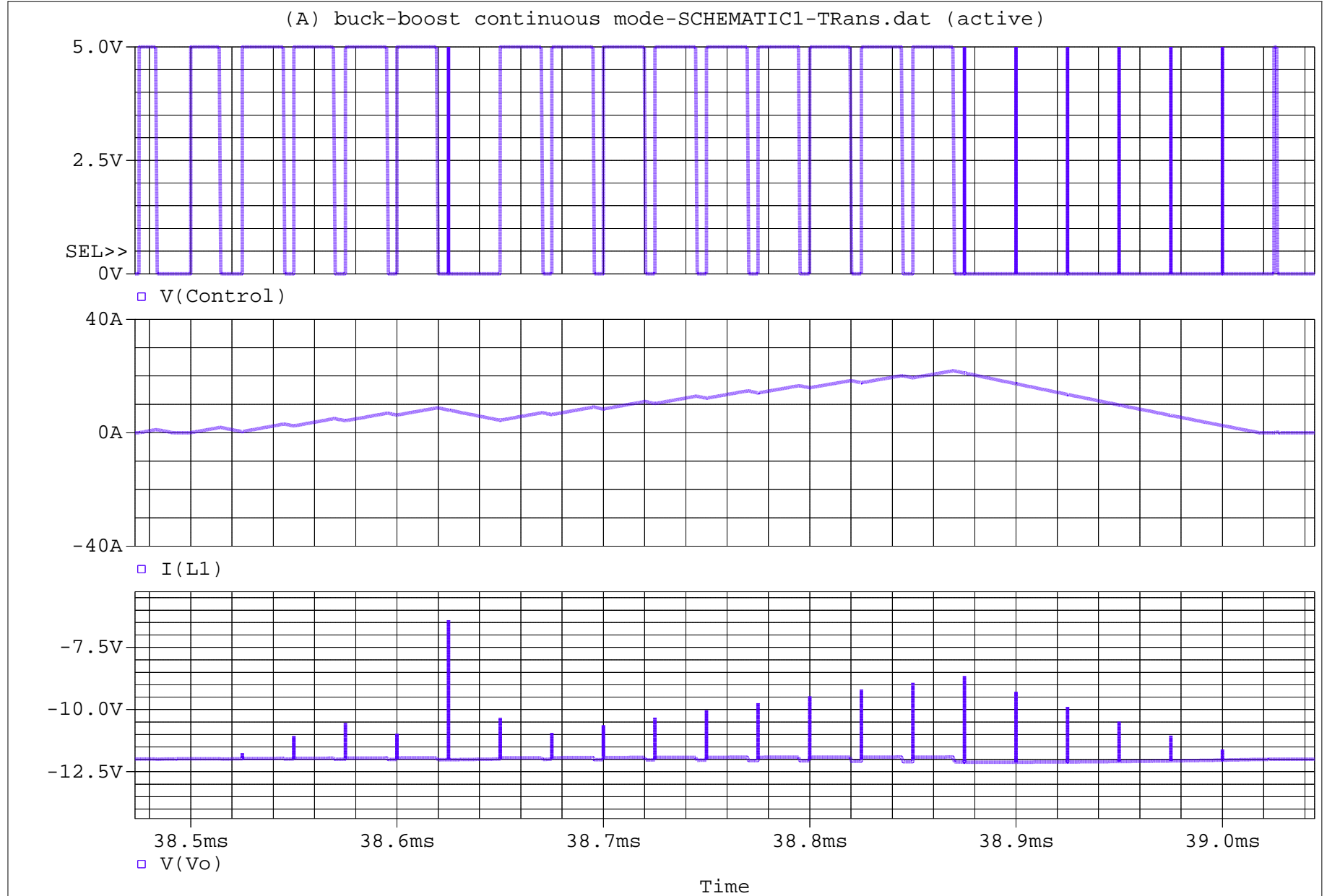
**ROSE-HULMAN**  
INSTITUTE OF TECHNOLOGY

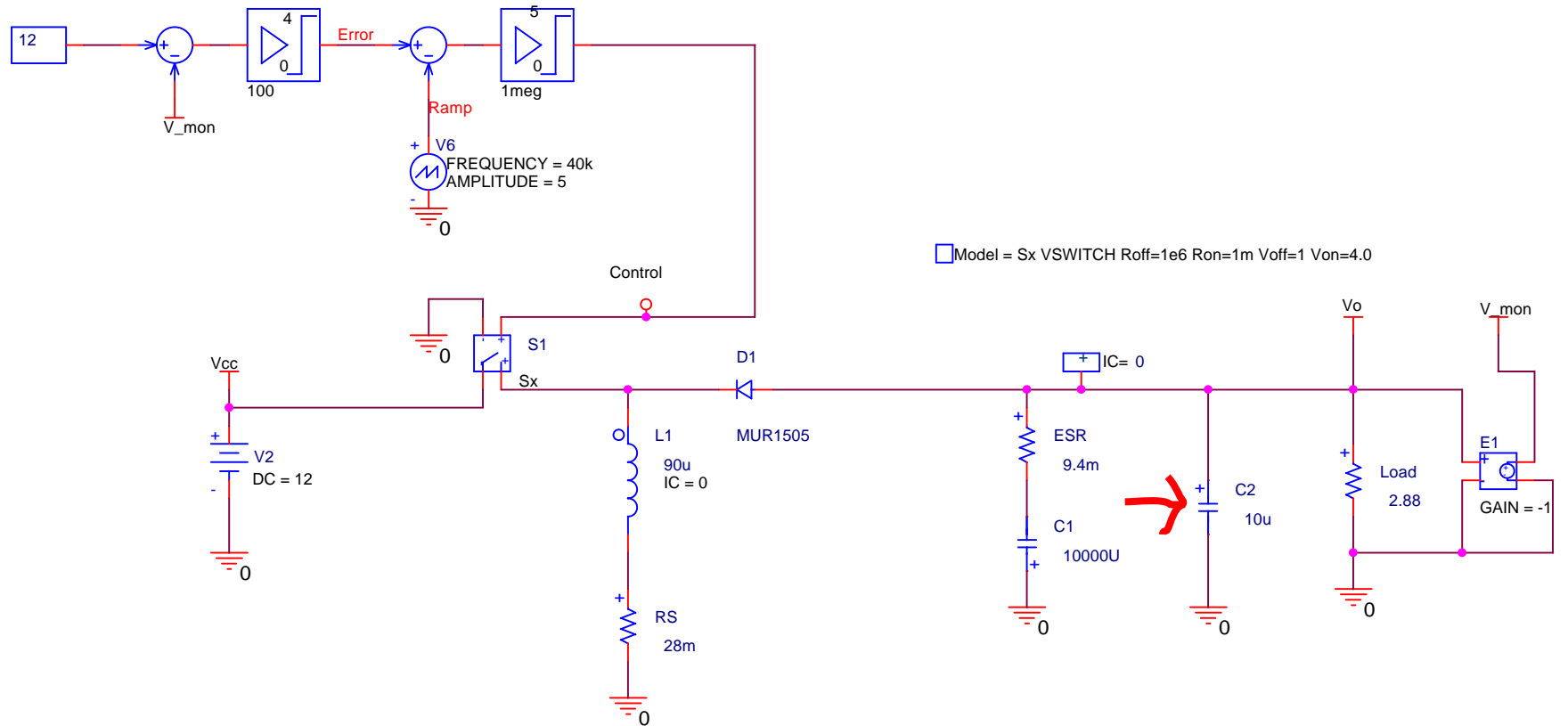
ECE Department  
5500 Wabash Avenue  
Terre Haute, IN 47803  
Ph: (812) 877-8512  
FAX: (253) 369-9536


Name: Marc E. Herniter		Class: ECE456	
Size A	Document Name		Rev 1
Date: Thursday, January 09, 2003		Sheet 1 of 1	

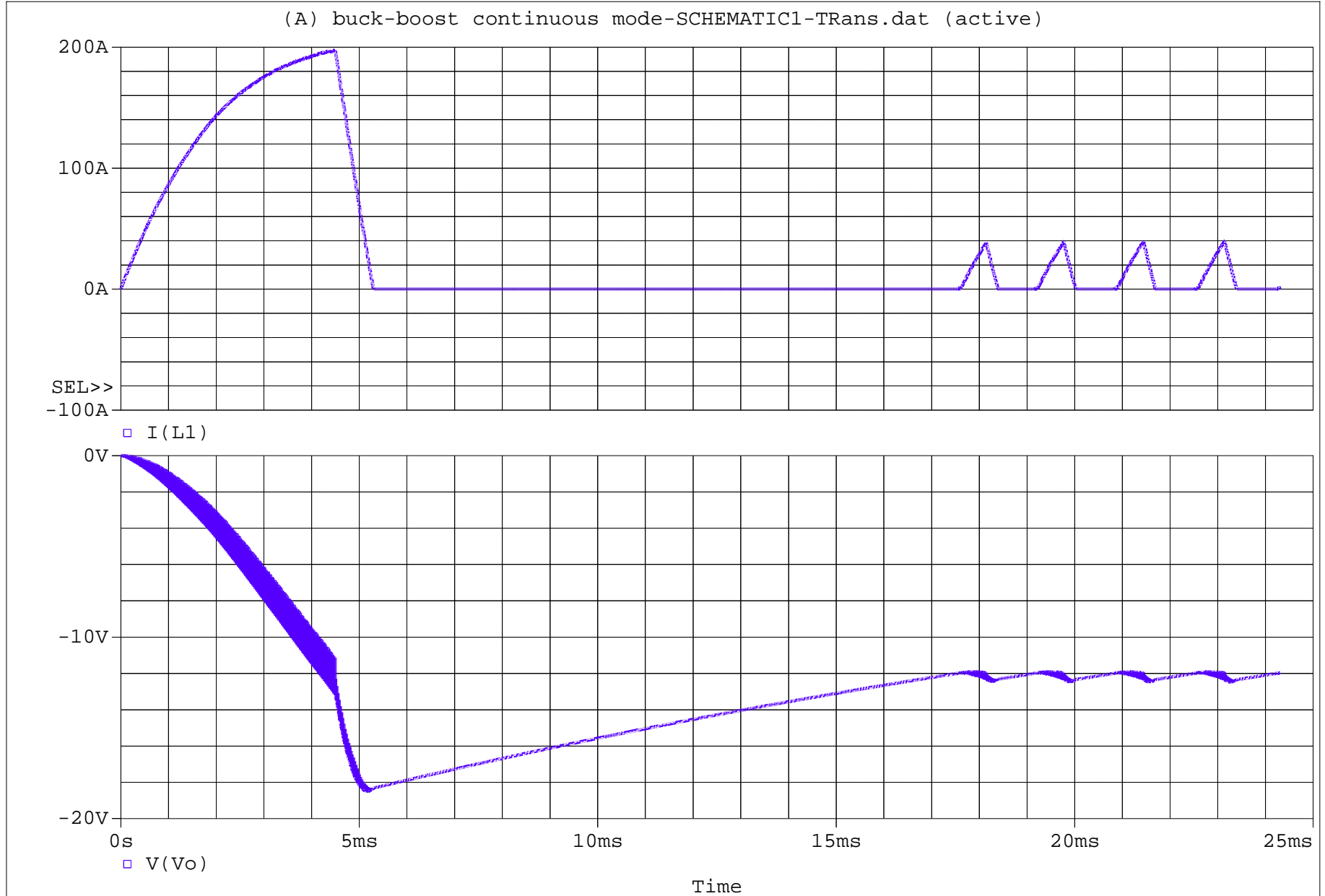


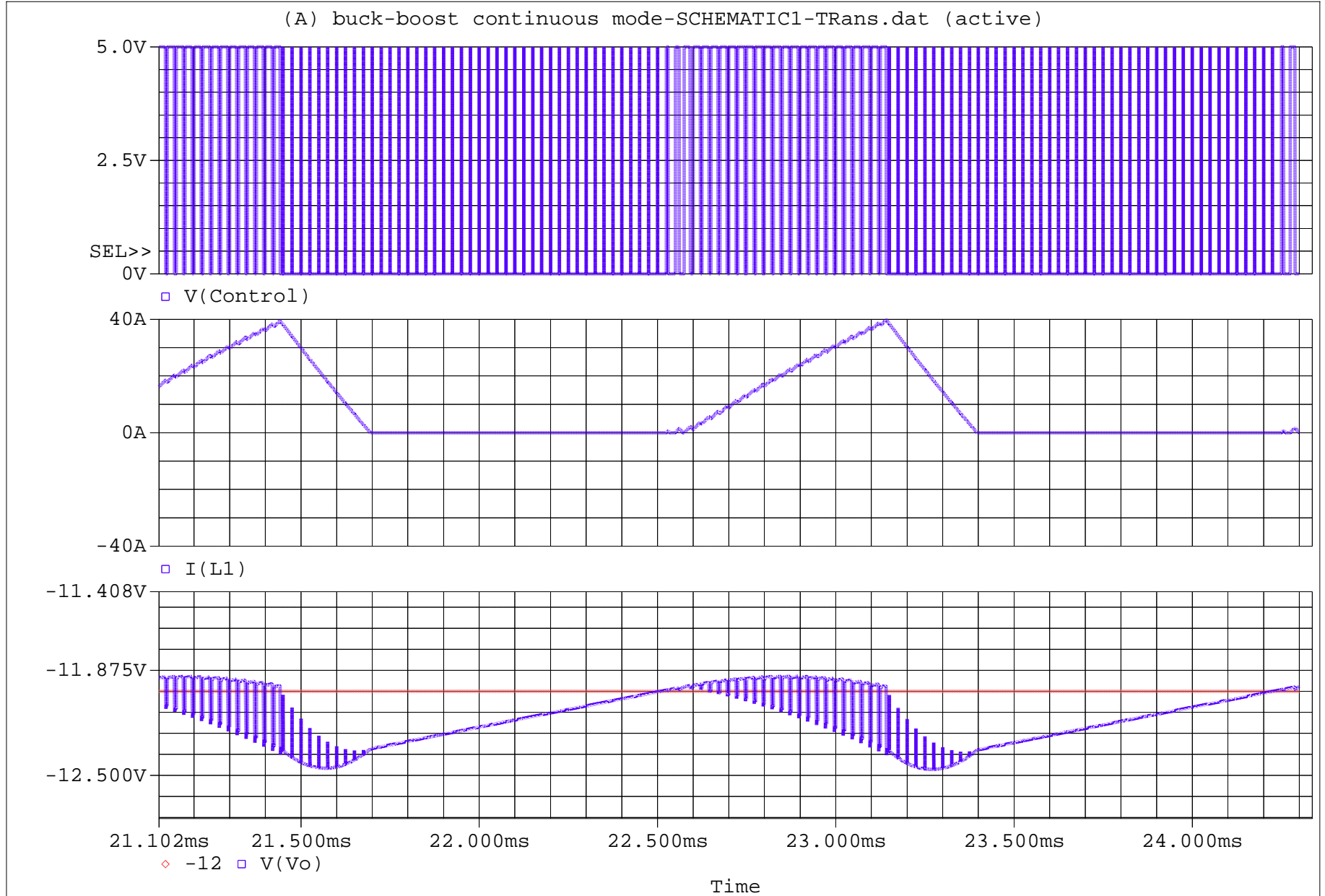


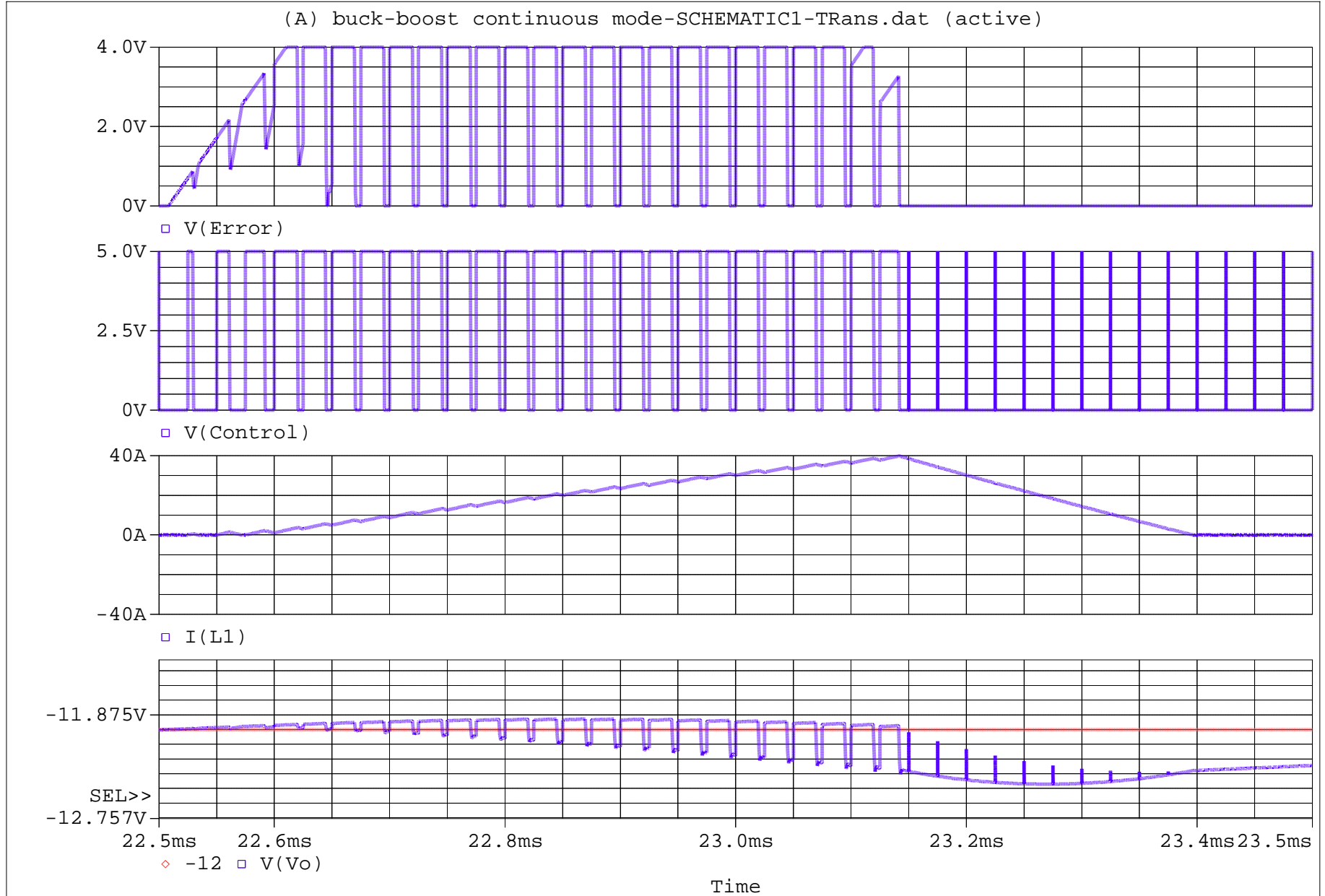


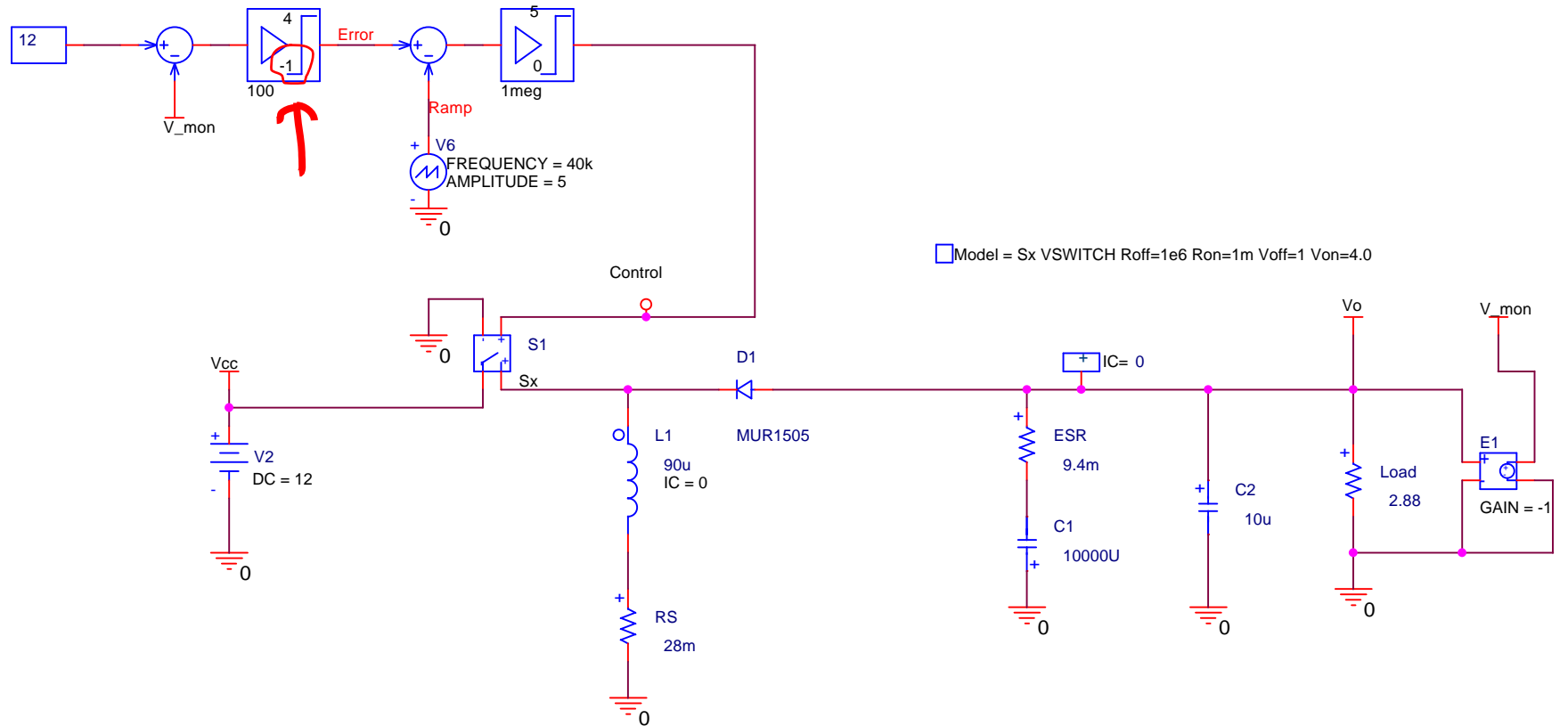



		ECE Department	
		5500 Wabash Avenue Terre Haute, IN 47803 Ph: (812) 877-8512 FAX: (253) 369-9536	
Name: Marc E. Herniter		Class: ECE456	
Size A	Document Name		Rev 1
Date: Thursday, January 09, 2003		Sheet 1 of 1	

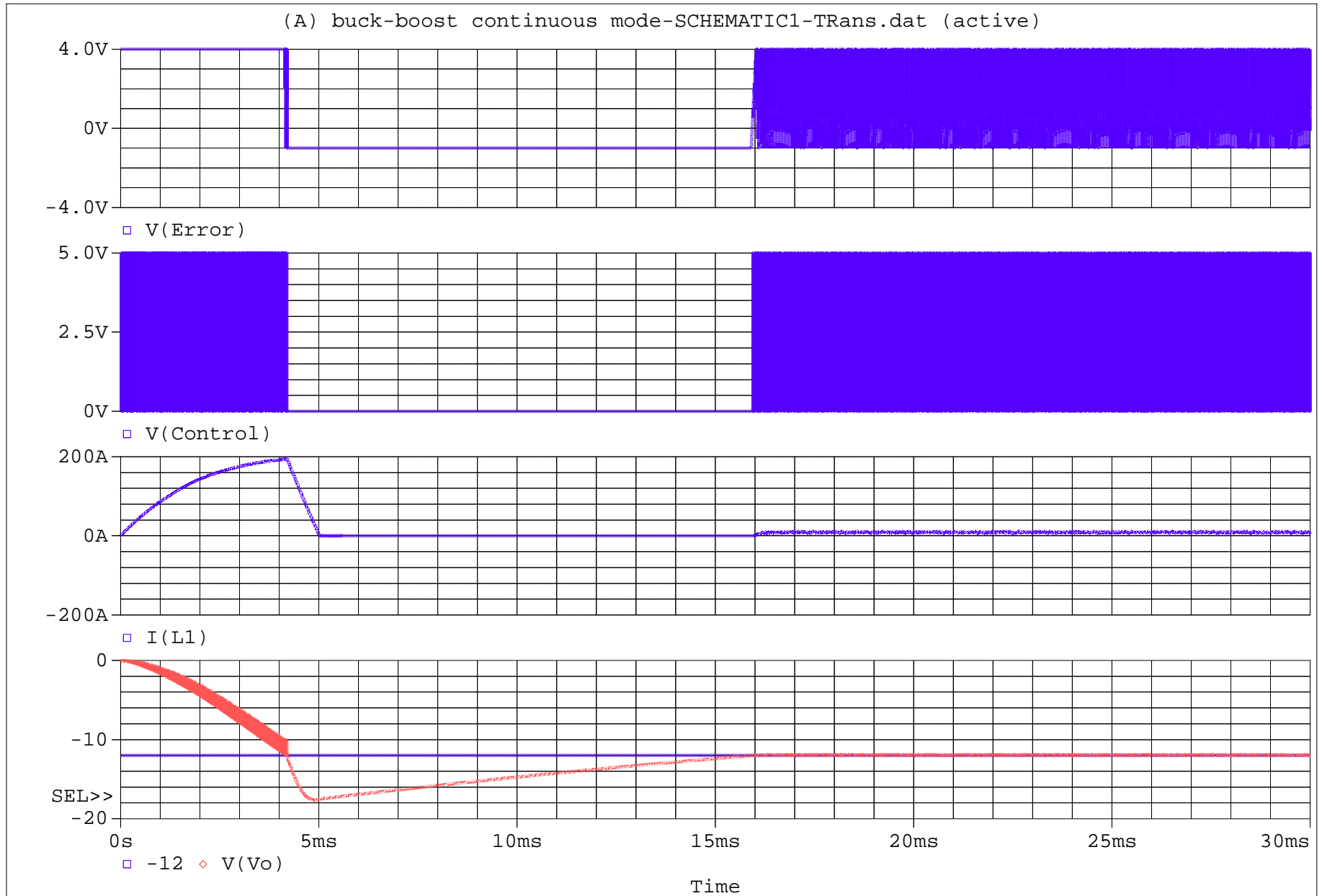


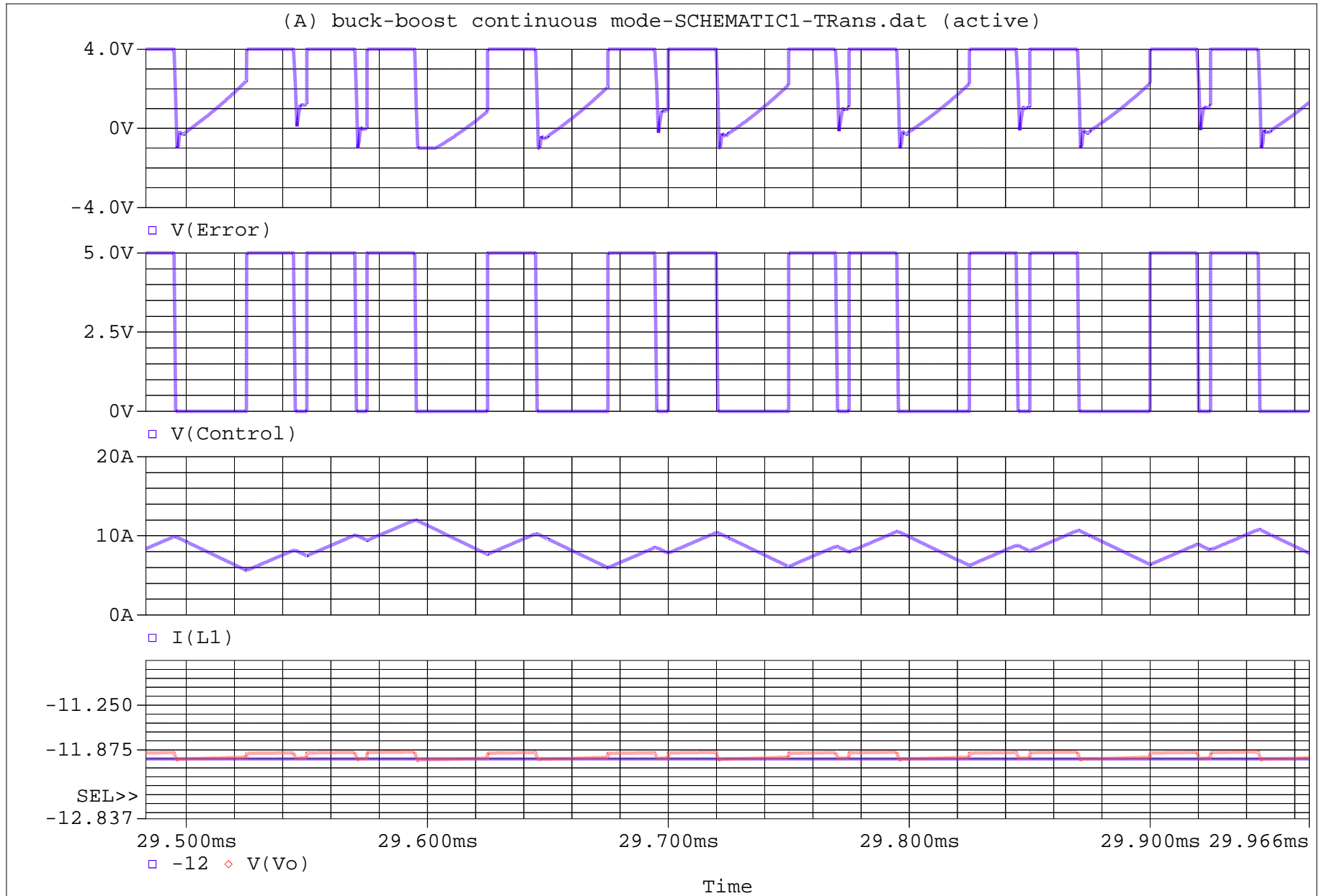


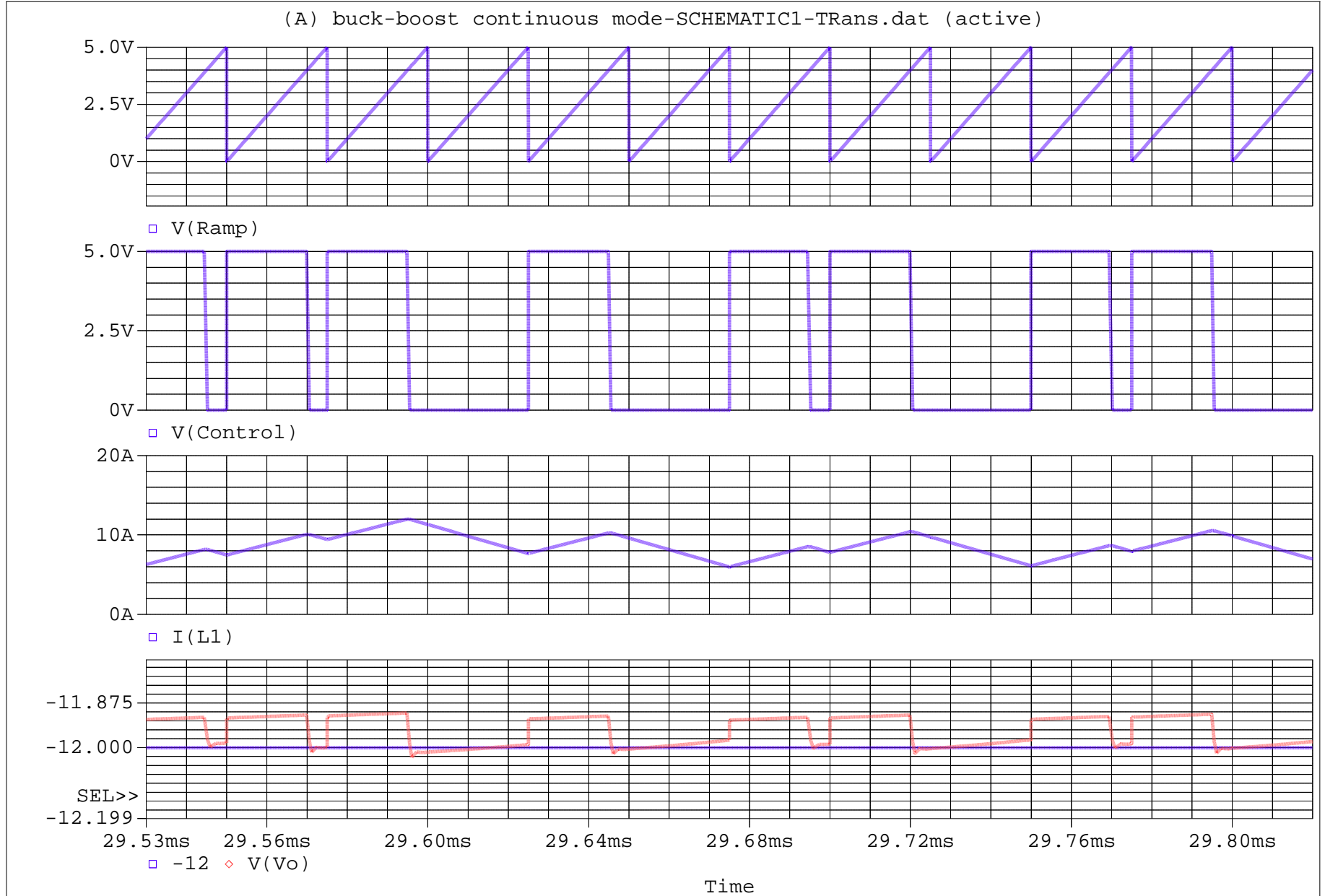


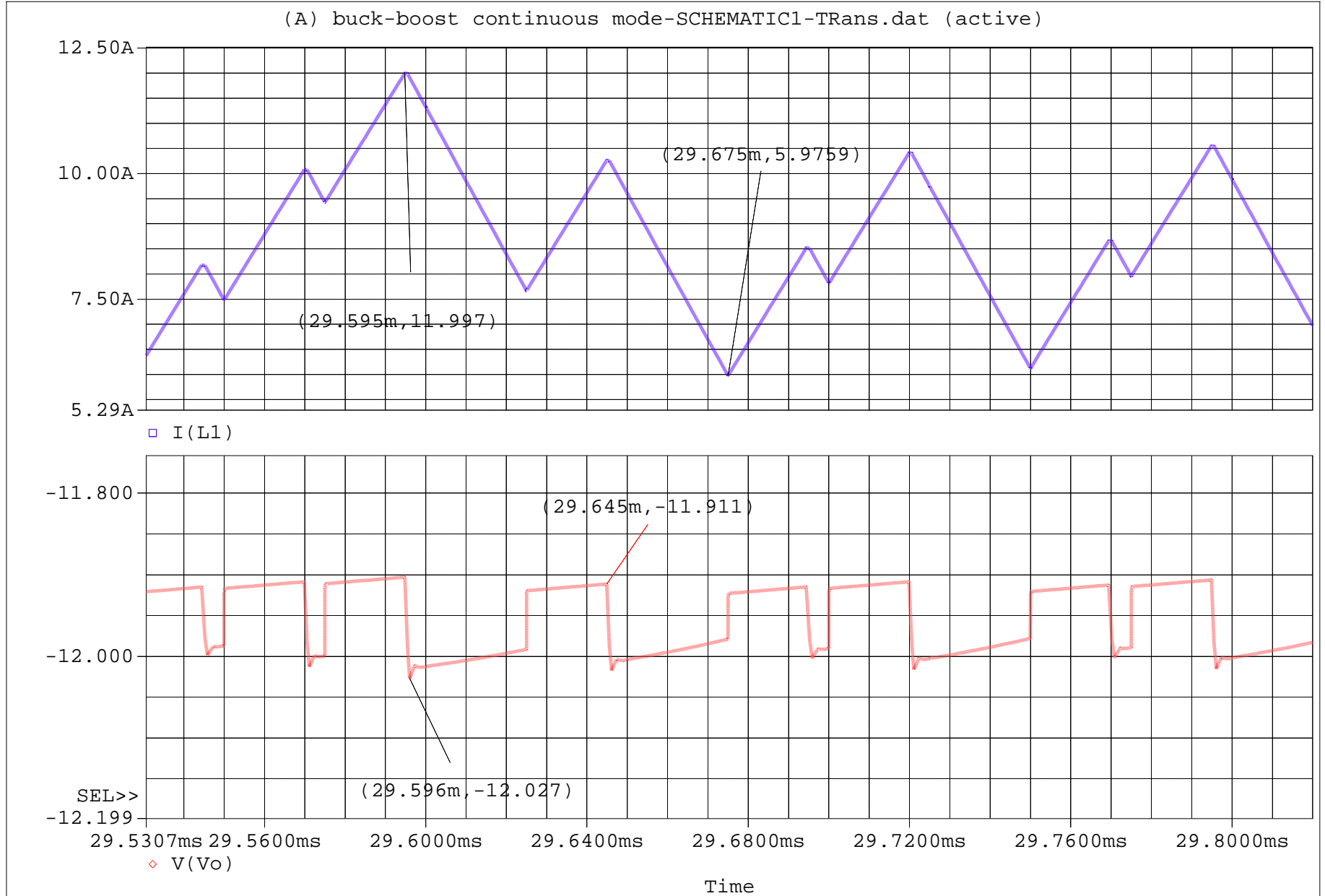


		ECE Department	
		5500 Wabash Avenue Terre Haute, IN 47803 Ph: (812) 877-8512 FAX: (253) 369-9536	
Name: Marc E. Herniter		Class: ECE456	
Size A	Document Name		Rev 1
Date: Thursday, January 09, 2003		Sheet 1 of 1	



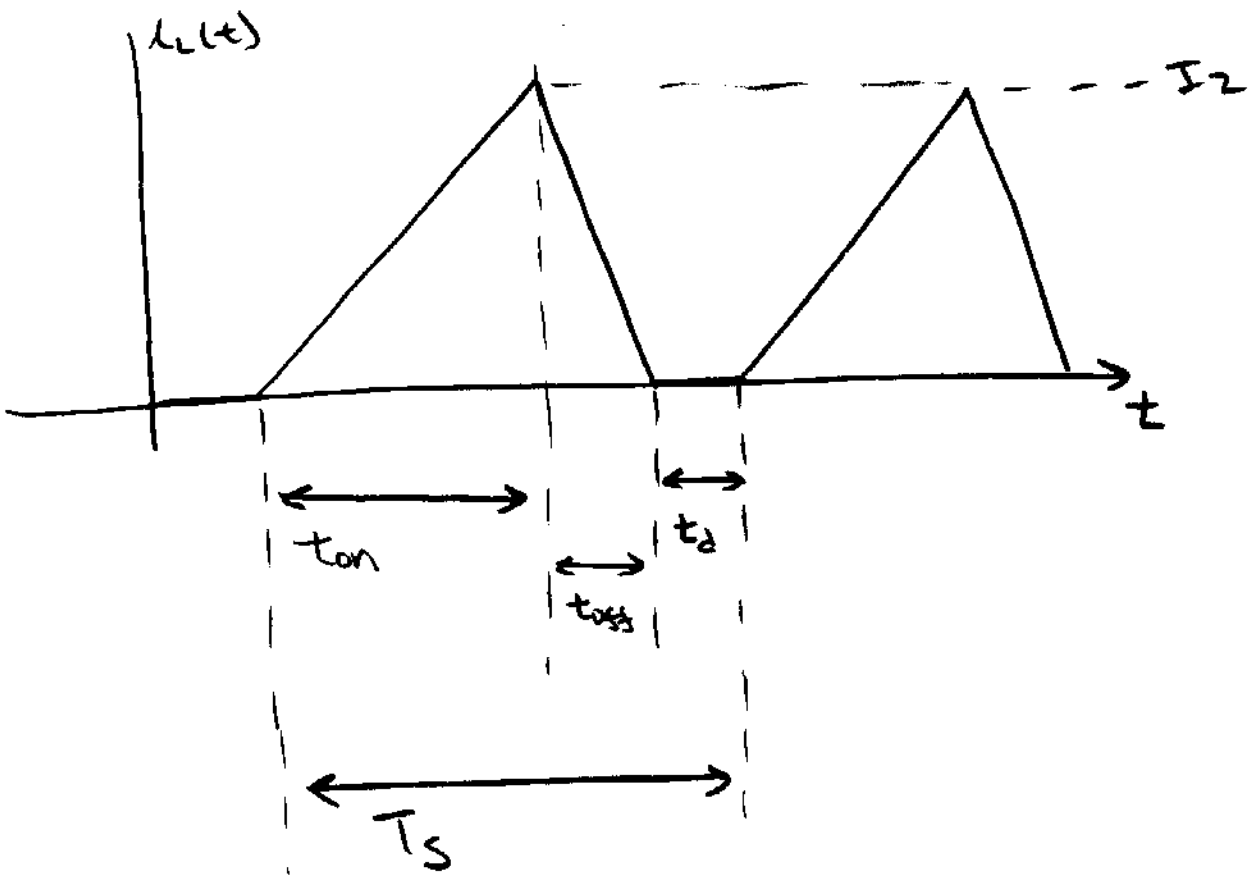
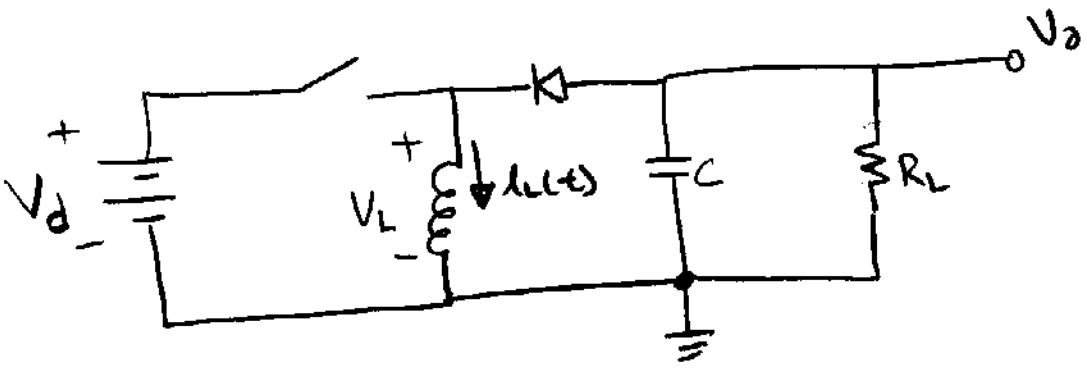




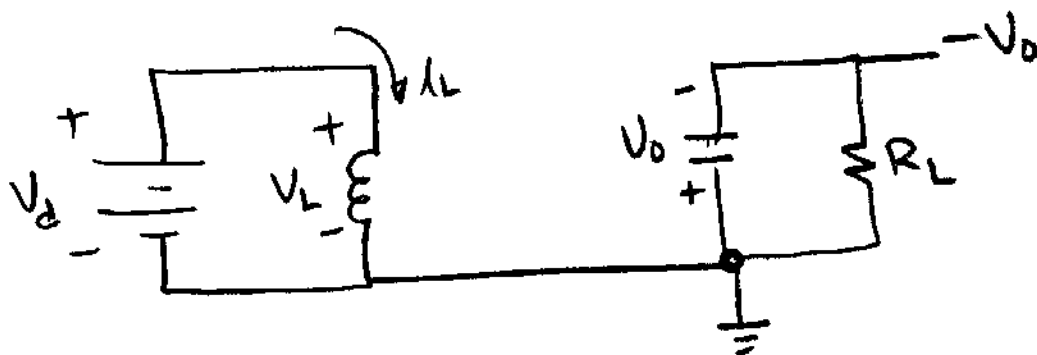


# BUCK-BOOST Regulator

Discontinuous mode



During  $t_{on}$  we have



For the inductor,  $V_L = L \frac{di_L}{dt}$

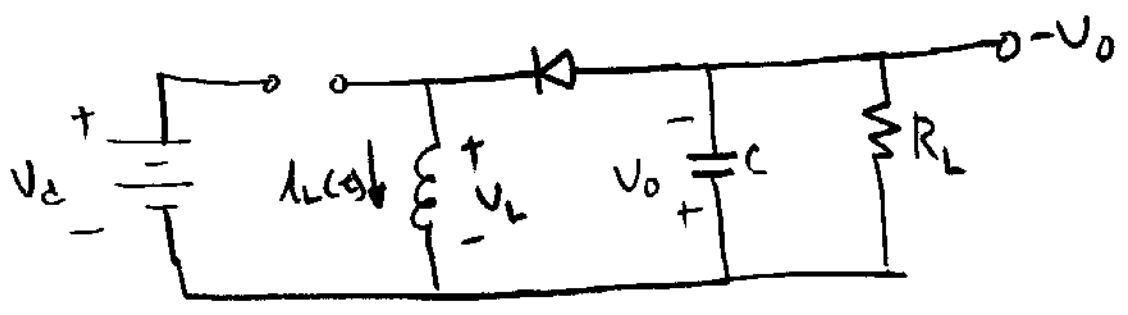
$$i_L(t) = \frac{1}{L} \int V_L(t) dt + i_{L.C.}$$

OR

$$I_2 = \frac{1}{L} \int_0^{t_{on}} V_d dt$$

$$I_2 = \frac{V_d t_{on}}{L} \quad \text{①}$$

During  $t_{off}$  We have



- Diode is Ideal
- D = ON
- $V_L = -V_o$

For the inductor  $V_L = L \frac{di_L}{dt}$

$$i_L(t) = \frac{1}{L} \int V_L(t) dt + I_2$$

OR

$$0 = \frac{1}{L} \int_{t_{on}}^{t_{on} + t_{off}} (-V_o) dt + I_2$$

$$I_2 = \frac{V_o t_{off}}{L} \quad \text{②}$$

- To ensure discontinuous mode operation

Choose  $t_d = 0.2 T_s$

$\Rightarrow t_{on} + t_{off} \leq 0.8 T_s$  (3)

Energy balance

- Energy dissipated by load during a cycle

$$E_o = V_o I_o T_s$$

- During  $t_{off}$ , all energy stored in L is delivered to Load

$$E_L = \frac{1}{2} L I_2^2$$

but from (1),  $I_2 = \frac{V_d t_{on}}{L}$

So

$$E_L = \frac{1}{2} L \left( \frac{V_d^2 t_{on}^2}{L^2} \right)$$

$$E_L = \frac{V_d^2 t_{on}^2}{2L}$$

Energy balance

$$E_L = E_o$$

$$\frac{V_d^2 t_{on}^2}{2L} = V_o I_o T_s$$

$$V_o = \frac{V_d^2 t_{on}^2}{2L I_o T_s}$$

 (4)

## BUCK-BOOST Summary

Discontinuous mode

$$\left. \begin{aligned} I_2 &= \frac{V_d t_{on}}{L} \\ I_2 &= \frac{V_o t_{off}}{L} \end{aligned} \right\} \Rightarrow \frac{t_{on}}{t_{off}} = \frac{V_o}{V_d}$$

$$t_{on} + t_{off} \leq 0.8 T_s$$

$$V_o = \frac{V_d^2 t_{on}^2}{2L I_o T_s}$$

$$I_o \leq \frac{V_d t_{on}}{2L} (1-D) \quad \text{For discontinuous mode}$$

# EE 456

## Buck-Boost Regulator Design Discontinuous Mode Operation

$$\text{m}\Omega \equiv \Omega \cdot 0.001 \quad \mu\text{s} := \text{sec} \cdot 10^{-6} \quad \mu\text{J} := \text{joule} \cdot 10^{-6}$$

Specify Input Voltage  $V_D := 12 \cdot \text{volt}$

Specify Output Voltage  $V_o := -12 \cdot \text{volt}$

Specify Switching Frequency  $F_S := 40 \cdot \text{kHz}$

$$T_S := \frac{1}{F_S} \quad T_S = 25 \mu\text{s}$$

Specify the Assumed Efficiency  $\text{Eff} := 85\%$

The output Power is  $P_{\text{out}} := \frac{50 \cdot \text{watt}}{\text{Eff}}$

The output current is  $I_o := \left| \frac{P_{\text{out}}}{V_o} \right| \quad I_o = 4.902 \text{ amp}$

Find Ton and Toff

$$t_{\text{off}} := 1 \cdot \mu\text{s} \quad t_{\text{on}} := 1 \cdot \mu\text{s}$$

Given

$$\frac{t_{\text{on}}}{t_{\text{off}}} = \frac{|V_o|}{V_D}$$

$$t_{\text{on}} + t_{\text{off}} = 0.8 \cdot T_S$$

$$\begin{pmatrix} t_{\text{on}} \\ t_{\text{off}} \end{pmatrix} := \text{Find}(t_{\text{on}}, t_{\text{off}}) \quad t_{\text{on}} = 10 \mu\text{s} \quad t_{\text{off}} = 10 \mu\text{s}$$

Find the range of Inductors that will operate in discontinuous mode

$$D := \frac{t_{\text{on}}}{T_S} \quad L := \frac{V_D \cdot t_{\text{on}}}{2 \cdot I_o} \cdot (1 - D)$$

For discontinuous Mode, We need L less than  $L = 7.344 \mu\text{H}$

Solve for the exact size inductor needed. Use the energy balance equation.

$$L := \frac{V_D^2 \cdot t_{\text{on}}^2}{2 \cdot |V_o| \cdot I_o \cdot T_S} \quad L = 4.896 \mu\text{H}$$

Find the peak current

$$I_2 := V_D \cdot \frac{t_{on}}{L} \quad I_2 = 24.51 \text{ amp}$$

Now that we know the approximate values needed for the inductor, choose a standard size inductor and then resolve the problem.

Some Pulse Engineering Specs    PE-51590     $L := 14 \cdot \mu\text{H}$

$$\text{Volt\_Sec} := 95 \cdot \text{volt} \cdot \mu\text{s} \quad \text{Energy\_Storage} := 700 \cdot \mu\text{J}$$

$$I_{DC} := 10 \cdot \text{amp}$$

$$I_{\max} := \frac{\text{Volt\_Sec}}{L} \quad I_{\max} = 6.786 \text{ amp}$$

$$I_{\max} := \sqrt{\frac{2 \cdot \text{Energy\_Storage}}{L}} \quad I_{\max} = 10 \text{ amp}$$

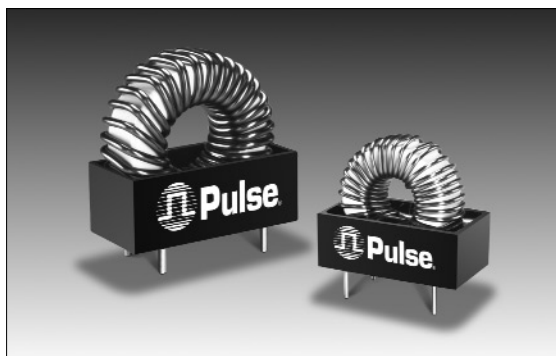
What is the saturation current of this inductor? I don't know. Use 2 in parallel to get  $7 \mu\text{H}$ . This will give us a different  $I_2$ ,  $t_{on}$ ,  $t_{off}$ .

$$L := 7 \cdot \mu\text{H}$$

Calculate the needed  $I_2$  to store the amount of energy needed per cycle.

$$I_2 := \sqrt{\frac{-2 \cdot V_o \cdot I_o \cdot T_S}{L}} \quad I_2 = 20.498 \text{ amp}$$

# TOROIDAL INDUCTORS HIGH CURRENT



- Cost-effective designs
- Semi-encapsulated construction
- Maximum operation temperature of 130°C (Ambient + Rise)
- A 2:1 inductance swing from zero to maximum current

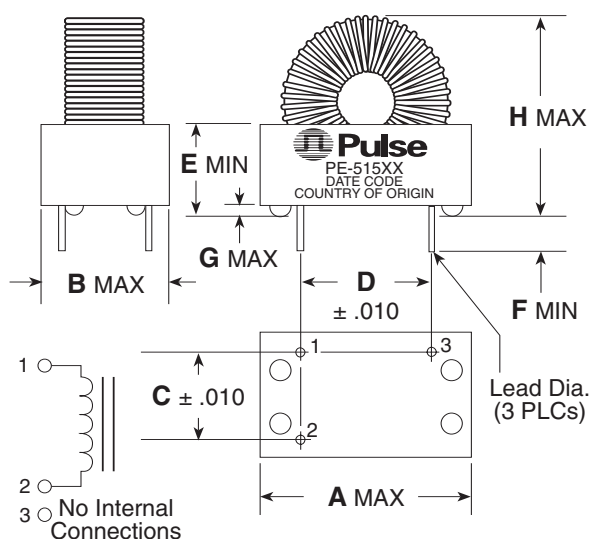
## Electrical Specifications @ 25°C

Part Number	REFERENCE OPERATING VALUES					DESIGN CONTROL VALUES				
	Inductance Typical (μH) <sup>2</sup>	I <sub>DC</sub> (AMPS)	ET <sub>OP</sub> <sup>1</sup> (V-μSec)		Energy Storage (μJ MIN) <sup>3</sup>	Inductance No DC (μH) (±20%)	50 kHz Test mV No DC <sub>s</sub>	DCR (Ω MAX)	Size Code	Lead Diameter (in ±.003)
			20 kHz	40 kHz						
PE-51506	17.0	17.0	190	130	2460	40.0	140	0.0065	3	0.081
PE-51507	32.0	16.0	290	200	4100	70.7	270	0.0092	4	0.081
PE-51508	60.0	16.0	390	270	7700	120.0	470	0.012	5	0.081
PE-51509	14.0	10.0	135	95	700	28.5	73	0.009	1	0.057
PE-51510	23.0	11.0	170	120	1400	43.3	130	0.012	2	0.057
PE-51511	43.0	10.0	280	195	2150	85.5	210	0.018	3	0.057
PE-51512	90.0	10.0	430	300	4500	158.0	420	0.028	4	0.057
PE-51513	144.0	10.0	570	400	7200	262.0	700	0.032	5	0.057
PE-51514	32.0	6.6	200	140	700	60.5	110	0.025	1	0.040
PE-51515	52.0	7.0	230	160	1275	92.0	190	0.032	2	0.040
PE-51516	98.0	6.0	400	280	1765	188.0	310	0.048	3	0.040
PE-51517	175.0	6.0	620	425	3150	315.0	560	0.068	4	0.040
PE-51518	335.0	6.0	840	580	6030	571.0	1000	0.095	5	0.040
PE-51520	400	3.6	600	420	2700	688.0	640	0.130	3	0.036

### NOTES:

- To prevent excessive temperature rise, limit ET<sub>OP</sub> to the rated ET<sub>OP</sub> specified. This is not a saturation limit. Temperature rise of inductors is 40°C MAX at MAX current and rated ET<sub>OP</sub>.
- A 2:1 nominal inductance swing from no I<sub>DC</sub> to operating I<sub>DC</sub> gives improved protection against current discontinuities at light loading. Inductance increases with greater ET<sub>OP</sub>. Reference values occur at I<sub>DC</sub> and low flux density.
- $\frac{LI^2}{2}$  rating is the ability of the inductor to store energy.
- Design control test voltage is critical. Inductance increases with voltage.

### Mechanicals



Size Code	1	2	3	4	5
A	1.20/30,48	1.44/36,57	1.60/40,64	1.95/49,53	2.30/58,42
B	0.60/15,24	0.80/20,32	0.80/20,32	0.91/23,11	1.11/28,19
C	0.40/10,16	0.60/15,24	0.60/15,24	0.70/17,78	0.90/22,85
D	0.80/20,32	0.90/22,86	0.90/22,86	1.20/30,48	1.50/38,10
E	0.45/11,43	0.70/17,78	0.70/17,78	0.90/22,86	1.00/25,40
F	0.20/5,08	0.20/5,08	0.20/5,08	0.20/5,08	0.20/5,08
G	.015/0,381	0.03/0,76	0.03/0,76	0.03/0,76	0.03/0,76
H	1.20/30,48	1.44/36,57	1.72/43,68	2.00/50,80	2.30/58,42

Dimensions:  $\frac{\text{Inches}}{\text{mm}}$

Unless otherwise specified, all tolerances are  $\pm \frac{.010}{.025}$

Now calculate the value of  $t_{\text{on}}$  and  $t_{\text{off}}$  to get that peak current.

$$t_{\text{on}} := \frac{I_2 \cdot L}{V_D} \qquad t_{\text{on}} = 11.957 \mu\text{s}$$

$$t_{\text{off}} := \frac{I_2 \cdot L}{-V_o} \qquad t_{\text{off}} = 11.957 \mu\text{s}$$

If  $t_{\text{on}} + t_{\text{off}} > T_S$  then the inductor is too big and the regulator will operate in the continuous mode.

$$\frac{t_{\text{on}} + t_{\text{off}}}{T_S} = 0.957$$

Calculate the dead time

$$t_d := T_S - t_{\text{on}} - t_{\text{off}} \qquad t_d = 1.086 \mu\text{s}$$

This dead time is too small. Try another inductor.

# EE 456

## Buck-Boost Regulator Design Discontinuous Mode Operation

$$\text{m}\Omega \equiv \Omega \cdot 0.001 \quad \mu\text{s} := \text{sec} \cdot 10^{-6} \quad \mu\text{J} := \text{joule} \cdot 10^{-6}$$

Specify Input Voltage  $V_D := 12 \cdot \text{volt}$

Specify Output Voltage  $V_o := -12 \cdot \text{volt}$

Specify Switching Frequency  $F_S := 40 \cdot \text{kHz}$

$$T_S := \frac{1}{F_S} \quad T_S = 25 \mu\text{s}$$

Specify the Assumed Efficiency  $\text{Eff} := 85\%$

The output Power is  $P_{\text{out}} := \frac{50 \cdot \text{watt}}{\text{Eff}}$

The output current is  $I_o := \left| \frac{P_{\text{out}}}{V_o} \right| \quad I_o = 4.902 \text{ amp}$

Find Ton and Toff

$$t_{\text{off}} := 1 \cdot \mu\text{s} \quad t_{\text{on}} := 1 \cdot \mu\text{s}$$

Given

$$\frac{t_{\text{on}}}{t_{\text{off}}} = \frac{|V_o|}{V_D}$$

$$t_{\text{on}} + t_{\text{off}} = 0.8 \cdot T_S$$

$$\begin{pmatrix} t_{\text{on}} \\ t_{\text{off}} \end{pmatrix} := \text{Find}(t_{\text{on}}, t_{\text{off}}) \quad t_{\text{on}} = 10 \mu\text{s} \quad t_{\text{off}} = 10 \mu\text{s}$$

Find the range of Inductors that will operate in discontinuous mode

$$D := \frac{t_{\text{on}}}{T_S} \quad L := \frac{V_D \cdot t_{\text{on}}}{2 \cdot I_o} \cdot (1 - D)$$

For discontinuous Mode, We need L less than  $L = 7.344 \mu\text{H}$

Solve for the exact size inductor needed. Use the energy balance equation.

$$L := \frac{V_D^2 \cdot t_{\text{on}}^2}{2 \cdot |V_o| \cdot I_o \cdot T_S} \quad L = 4.896 \mu\text{H}$$

Find the peak current

$$I_2 := V_D \cdot \frac{t_{on}}{L} \quad I_2 = 24.51 \text{ amp}$$

Now that we know the approximate values needed for the inductor, choose a standard size inductor and then resolve the problem.

Some Pulse Engineering Specs    PE-51590     $L := 22 \cdot \mu\text{H}$

$$\text{Volt\_Sec} := 44 \cdot \text{volt} \cdot \mu\text{s} \quad \text{Energy\_Storage} := 275 \cdot \mu\text{J}$$

$$I_{DC} := 5 \cdot \text{amp}$$

$$I_{\max} := \frac{\text{Volt\_Sec}}{L} \quad I_{\max} = 2 \text{ amp}$$

$$I_{\max} := \sqrt{\frac{2 \cdot \text{Energy\_Storage}}{L}} \quad I_{\max} = 5 \text{ amp}$$

What is the saturation current of this inductor? I don't know. Use 4 in parallel to get  $5.5 \mu\text{H}$ . This will give us a different  $I_2$ ,  $t_{on}$ ,  $t_{off}$ .

Coil Craft Inductor PCV-0-103-20     $L := 10 \cdot \mu\text{H}$

$$I_{SAT} := 20 \cdot \text{amp}$$






Use two of the coilcraft Parts in parallel.

$$L := 5 \cdot \mu\text{H}$$

# LOW COST INDUCTORS

## Electrical Information



-  Available in vertical, low profile and *KlipMount™*
-  SMPS averaging filter
-  Characterized for general purpose use and ripple filters
-  Single-layer designs
-  Can be used as differential mode inductors in EMI filters<sup>3</sup>

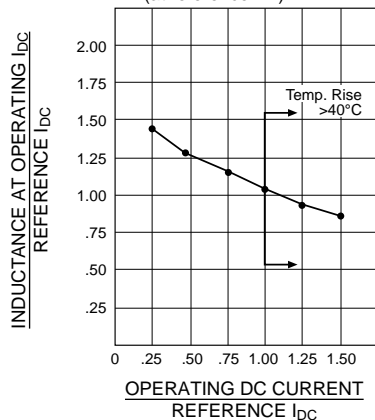
### Electrical Specifications @ 25°C — Operating Temperature -30°C to 130°C

REFERENCE OPERATING VALUES						DESIGN CONTROL VALUES					
Vertical Part Number	Low Profile Part Number	Inductance Typical (μH) <sup>1</sup>	I <sub>DC</sub> (AMPS)	ET <sub>TOP</sub> (V-μSec)	Energy Storage (μJ) <sup>4</sup>	Inductance No DC (μHy) ±20%	20 KHz Test mV No DC <sup>2</sup>	DCR (Ω MAX) <sup>6</sup>	Coil Size Code	Klip Mount Package*	Lead Diameter (In) ±.003
PE-51591	PE-92421	20	2.0	52	40	32.8	33	.060	H	—	.020
PE-92100	—	25	2.6	30	85	20.7	22	.043	A	KM1	.020
PE-92101	PE-92401	50	2.6	50	169	45.7	45	.071	B	KM2	.020
PE-92102	PE-92402	100	2.6	90	338	94.1	90	.100	C	KM3	.020
PE-92103	—	35	2.6	55	118	28.4	36	.037	B	KM2	.025
PE-92104	PE-92404	70	3.0	85	315	61.0	73	.052	C	KM3	.025
PE-92105	PE-92405	145	3.0	140	653	141.8	140	.087	D	KM4	.025
PE-92106	—	285	3.0	300	1283	264.1	340	.140	E	KM5	.025
PE-92107	—	450	3.0	425	2025	436.3	500	.200	F	—	.025
PE-92108	PE-92408	100	3.6	130	648	90.7	110	.045	D	KM4	.032
PE-92109	—	165	4.0	240	1320	152.0	260	.070	E	KM5	.032
PE-92110	—	270	4.0	350	2160	263.9	400	.100	F	—	.032
PE-92111	—	40	4.0	70	320	37.9	57	.027	C	KM3	.032
PE-51590	PE-92420	22	5.0	44	275	20.3	37	.020	G	—	.032
PE-92112	PE-92412	100	5.0	200	1250	90.7	180	.034	E	KM5	.042
PE-92113	—	170	5.0	300	2125	159.7	310	.050	F	—	.042
PE-92114	PE-92414	55	5.0	100	688	54.9	88	.023	D	KM4	.042
PE-92115	—	95	7.0	225	2328	96.0	200	.025	F	—	.051
PE-92116	PE-92416	55	7.0	150	1348	49.1	100	.017	E	KM5	.051
PE-92117	—	55	10.0	175	2750	55.9	120	.013	F	—	.064

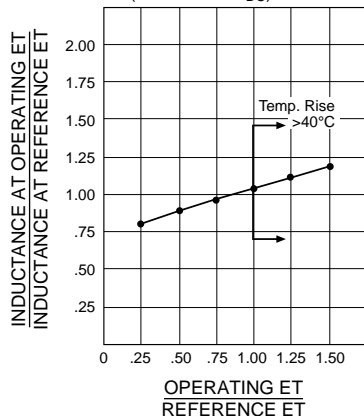
\*Parts available with *KlipMount* option can be ordered by adding a "K" suffix to the part number (i.e. PE-92100K).

### Relationships Between Reference and Operating Conditions

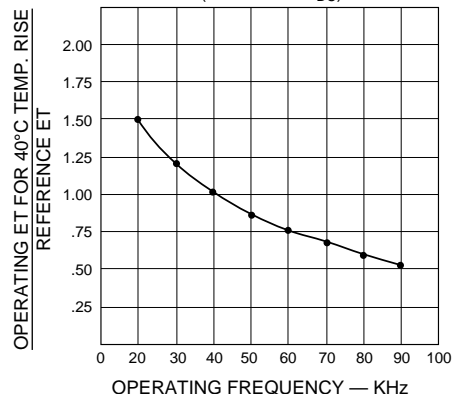
Inductance vs. DC Current  
(at reference ET)



Inductance vs. Operating ET  
(at reference I<sub>DC</sub>)



Max. Operating ET vs. Frequency  
(at reference I<sub>DC</sub>)



## PCV-0 Series – Continued

Part Number	Inductance <sup>1</sup> ( $\mu$ H)	Isat <sup>2</sup> (Amps)	Irms <sup>3</sup> (Amps)	DCR Max ( $\Omega$ )	A Max	B Max	C $\pm .015/0,38$	D $\pm .004/0,10$
PCV-0-103-20	10 $\pm 10\%$	20	20	.006	.75/19,1	1.8/45,7	.60/15,2	.075/1,91
PCV-0-153-03	15	3	3	.025	.50/12,7	1.0/25,4	.42/10,7	.035/0,89
PCV-0-153-10	15	10	10	.020	.52/13,2	1.0/25,4	.42/10,7	.054/1,37
<b>PCV-0-223-03</b>	<b>22</b>	<b>3</b>	<b>3</b>	<b>.035</b>	<b>.50/12,7</b>	<b>1.0/25,4</b>	<b>.42/10,7</b>	<b>.035/0,89</b>
<b>PCV-0-223-05</b>	<b>22</b>	<b>5</b>	<b>5</b>	<b>.023</b>	<b>.50/12,7</b>	<b>1.0/25,4</b>	<b>.42/10,7</b>	<b>.042/1,07</b>
<b>PCV-0-223-10</b>	<b>22</b>	<b>10</b>	<b>10</b>	<b>.015</b>	<b>.66/16,8</b>	<b>1.3/33,0</b>	<b>.42/10,7</b>	<b>.060/1,52</b>
PCV-0-273-05	27	5	5	.024	.50/12,7	1.0/25,4	.49/12,5	.042/1,07
PCV-0-333-12	33	12	12	.017	.70/17,8	1.3/33,0	.55/14,0	.060/1,52
<b>PCV-0-473-03</b>	<b>47</b>	<b>3</b>	<b>3</b>	<b>.050</b>	<b>.55/14,0</b>	<b>1.0/25,4</b>	<b>.42/10,7</b>	<b>.035/0,89</b>
<b>PCV-0-473-05</b>	<b>47</b>	<b>5</b>	<b>5</b>	<b>.035</b>	<b>.65/16,5</b>	<b>1.1/28,0</b>	<b>.70/17,8</b>	<b>.042/1,07</b>
<b>PCV-0-473-10</b>	<b>47</b>	<b>10</b>	<b>10</b>	<b>.022</b>	<b>.85/21,6</b>	<b>1.3/33,0</b>	<b>.70/17,8</b>	<b>.060/1,52</b>
PCV-0-823-03	82	3	3	.110	.50/12,7	.65/16,5	.375/9,5	.028/0,71
PCV-0-104-01	100	1	1	.190	.40/10,2	.90/22,9	.30/7,6	.020/0,51
<b>PCV-0-104-03</b>	<b>100</b>	<b>3</b>	<b>3</b>	<b>.072</b>	<b>.55/14,0</b>	<b>1.2/30,5</b>	<b>.70/17,8</b>	<b>.035/0,89</b>
<b>PCV-0-104-05</b>	<b>100</b>	<b>5</b>	<b>5</b>	<b>.055</b>	<b>.65/16,5</b>	<b>1.3/33,0</b>	<b>.70/17,8</b>	<b>.042/1,07</b>
PCV-0-154-03	150	3	3	.140	.60/15,2	1.2/30,5	.43/10,9	.028/0,71
PCV-0-154-05	150	5	5	.065	.65/16,5	1.3/33,0	.70/17,8	.042/1,07
PCV-0-184-05	180	5	5	.110	.60/15,2	1.2/30,5	.43/10,9	.035/0,89
PCV-0-224-03	220	3	3	.210	.55/14,0	1.2/30,5	.42/10,7	.025/0,64
PCV-0-274-04	270	4	4	.250	.95/24,0	.72/18,3	.71/18,0	.030/0,76
PCV-0-274-10	270	10	10	.160	1.1/28,0	1.0/25,4	.72/18,3	.038/0,97
PCV-0-394-03	390	3	3	.250	1.1/28,0	1.0/25,4	.72/18,3	.035/0,89
PCV-0-394-05	390	5	5	.190	1.1/28,0	1.0/25,4	.72/18,3	.038/0,97

## PCV-1 Series

Part Number	Inductance <sup>1</sup> ( $\mu$ H)	Isat <sup>2</sup> (Amps)	Irms <sup>3</sup> (Amps)	DCR Max ( $\Omega$ )	A Max	B Max	C $\pm .015/0,38$	D $\pm .004/0,10$
PCV-1-152-15	1.5 $\pm 15\%$	15	15	.003	.65/16,5	.90/22,9	.42/10,7	.068/1,73
PCV-1-182-10	1.8	10	10	.006	.65/16,5	1.1/27,9	.42/10,7	.054/1,37
PCV-1-182-15	1.8	15	15	.003	.65/16,5	1.1/27,9	.42/10,7	.068/1,73
PCV-1-472-03	4.7	3	3	.021	.62/15,7	.85/21,6	.42/10,7	.035/0,89
PCV-1-472-05	4.7	5	5	.012	.62/15,7	.85/21,6	.42/10,7	.042/1,07
PCV-1-472-10	4.7	10	10	.012	.62/15,7	1.1/27,9	.42/10,7	.054/1,37
PCV-1-103-03	10 $\pm 10\%$	3	3	.026	.63/16,0	.85/21,6	.42/10,7	.035/0,89
PCV-1-103-05	10	5	5	.020	.62/15,7	.85/21,6	.42/10,7	.042/1,07
PCV-1-103-10	10	10	10	.013	.63/16,0	1.1/27,9	.42/10,7	.054/1,37
PCV-1-153-10	15	10	10	.020	.70/17,8	1.4/35,6	.42/10,7	.060/1,52
PCV-1-223-03	22	3	3	.035	.62/15,7	1.1/27,9	.42/10,7	.035/0,89
PCV-1-223-05	22	5	5	.023	.62/15,7	1.1/27,9	.42/10,7	.042/1,07
PCV-1-223-10	22	10	10	.018	.70/17,8	1.4/35,6	.42/10,7	.060/1,52
PCV-1-473-03	47	3	3	.050	.65/16,5	1.1/27,9	.42/10,7	.035/0,89
PCV-1-473-05	47	5	5	.033	.90/22,9	1.2/30,5	.70/17,8	.042/1,07
PCV-1-473-10	47	10	10	.022	.90/22,9	1.4/35,6	.70/17,8	.060/1,52
PCV-1-104-03	100	3	3	.072	.90/22,9	1.3/33,1	.70/17,8	.035/0,89
PCV-1-104-05	100	5	5	.055	.90/22,9	1.4/35,6	.70/17,8	.042/1,07
PCV-1-184-03	180	3	3	.150	.63/16,0	1.2/30,5	.42/10,7	.028/0,71
PCV-1-304-05	300	5	5	.160	1.1/27,9	1.0/25,4	.70/17,8	.038/0,97
PCV-1-394-05	390	5	5	.190	1.1/27,9	1.0/25,4	.70/17,8	.038/0,97

<sup>1</sup> Tested at 15.75 kHz, .1 Vrms, with DC bias applied up to the rated current.

<sup>2</sup> Inductance drop = 10% typ. at Isat.

<sup>3</sup>  $\Delta T = 40^\circ\text{C}$  rise typ. at Irms (85°C ambient).

Parts in bold type are included in Coilcraft Designer's Kit No. P205.



Specifications subject to change without notice. Document 135-2 Revised 7/31/00

1102 Silver Lake Road Cary, Illinois 60013 Phone 847/639-6400 Fax 847/639-1469  
E-mail info@coilcraft.com Web http://www.coilcraft.com

Calculate the needed  $I_2$  to store the amount of energy needed per cycle.

$$I_2 := \sqrt{\frac{-2 \cdot V_o \cdot I_o \cdot T_S}{L}} \quad I_2 = 24.254 \text{ amp}$$

Now calculate the value of  $t_{\text{on}}$  and  $t_{\text{off}}$  to get that peak current.

$$t_{\text{on}} := \frac{I_2 \cdot L}{V_D} \quad t_{\text{on}} = 10.106 \mu\text{s}$$

$$t_{\text{off}} := \frac{I_2 \cdot L}{-V_o} \quad t_{\text{off}} = 10.106 \mu\text{s}$$

If  $t_{\text{on}} + t_{\text{off}} > T_S$  then the inductor is too big and the regulator will operate in the continuous mode.

$$\frac{t_{\text{on}} + t_{\text{off}}}{T_S} = 0.808$$

Calculate the dead time

$$t_d := T_S - t_{\text{on}} - t_{\text{off}} \quad t_d = 4.789 \mu\text{s}$$

This dead time is OK.

Choose the filter capacitor using the capacitor ESR.

Specify the ripple due to the ESR  $V_{CR} := 100 \cdot \text{mV}$

$$\text{ESR} := \frac{V_{CR}}{I_2} \quad \text{ESR} = 4.123 \text{ m}\Omega$$

For all electrolytic caps, assume that  $\text{ESR} \cdot C = 80 \text{ms}$

$$C := \frac{80 \cdot \mu\text{s}}{\text{ESR}} \quad C = 19403 \mu\text{F}$$

Choose the next size std capacitor  $C := 22000 \cdot \mu\text{F}$

Calculate the Capacitor RMS Ripple Current

Define a function for the inductor current  $I_L(t) := I_2 - \frac{I_2}{t_{\text{off}}} \cdot t$

$$I_{\text{rms}} := \sqrt{\frac{1}{T_S} \cdot \left[ \int_{0 \cdot \text{sec}}^{t_{\text{on}} + t_d} (-I_0)^2 dt + \int_{0 \cdot \text{sec}}^{t_{\text{off}}} (I_L(t) - I_0)^2 dt \right]}$$

$$I_{\text{rms}} = 7.432 \text{ amp}$$

Summary

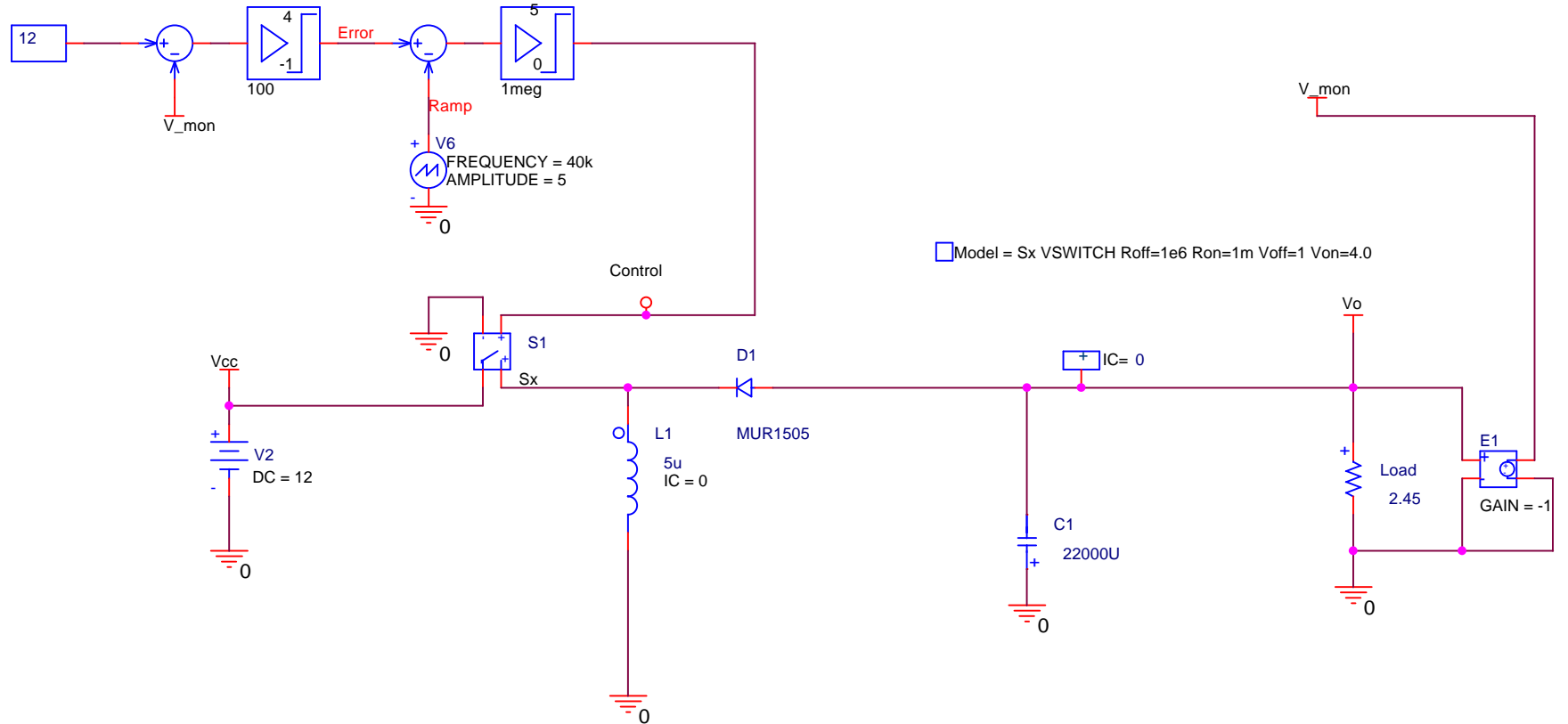
$$L = 5 \mu\text{H} \quad I_2 = 24.254 \text{ amp}$$


$$t_{\text{on}} = 10.106 \mu\text{s} \quad t_{\text{off}} = 10.106 \mu\text{s}$$

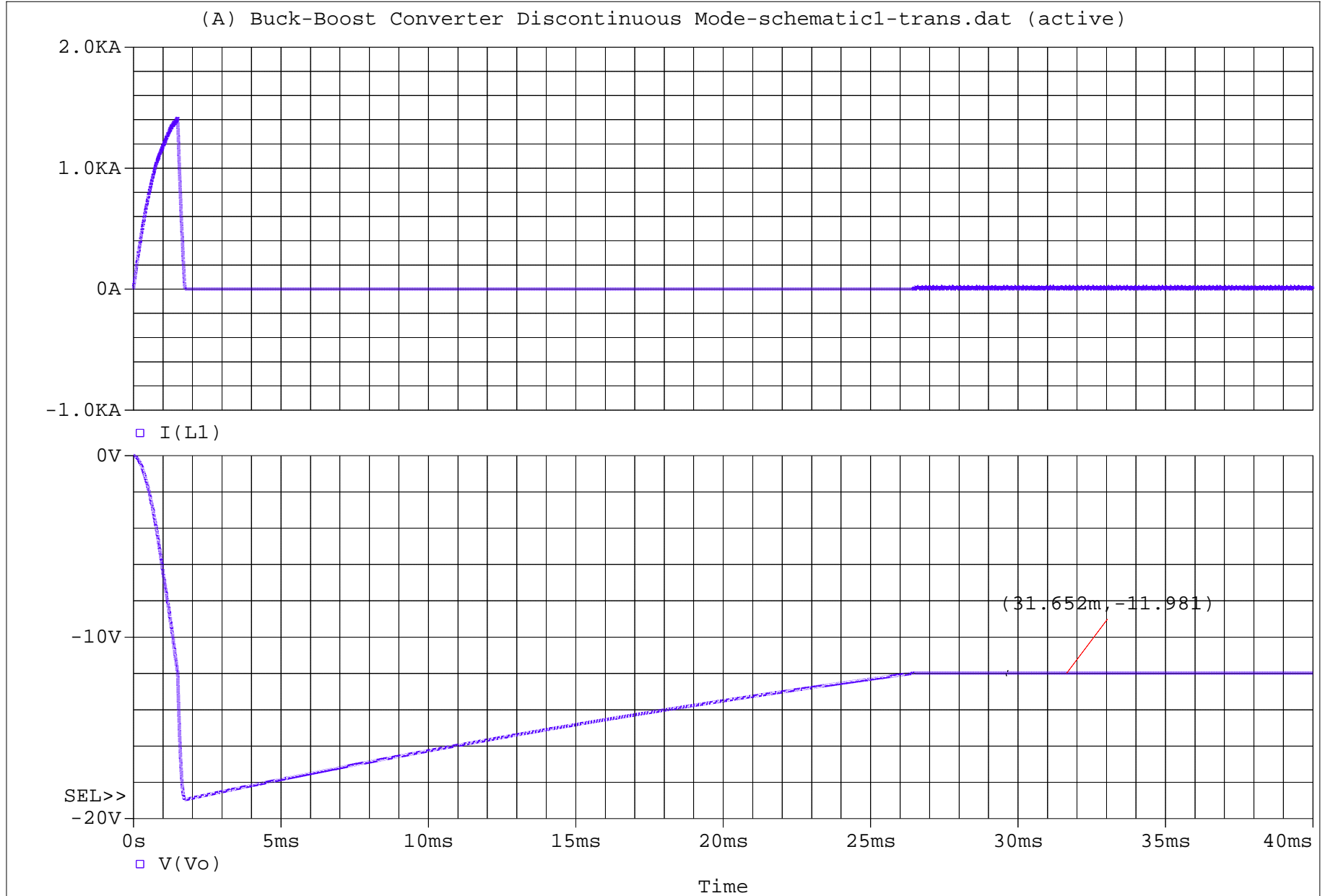
$$V_D = 12 \text{ volt} \quad V_o = -12 \text{ volt} \quad I_o = 4.902 \text{ amp}$$

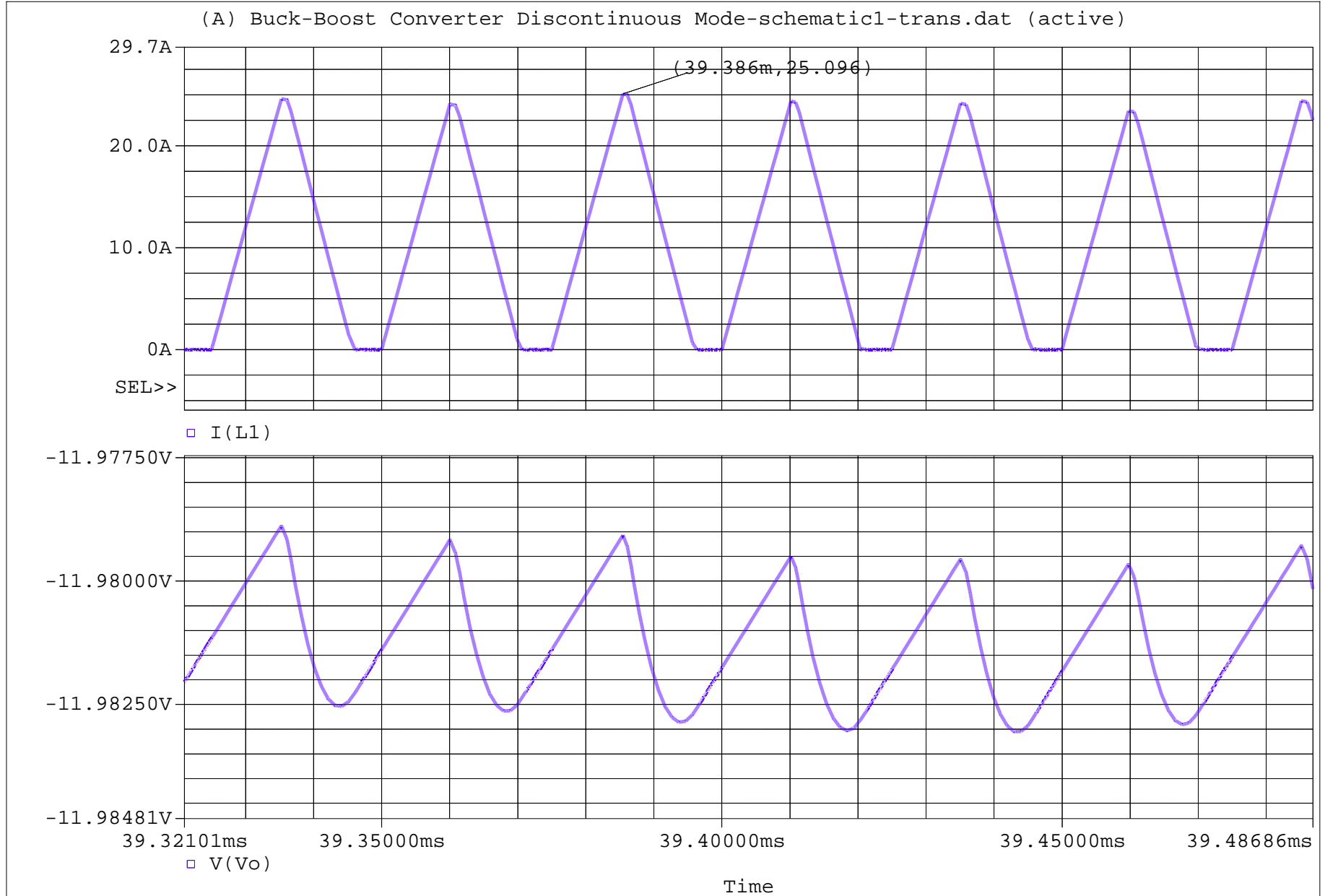
$$C = 22000 \mu\text{F} \quad V_{\text{CR}} = 100 \text{ mV}$$

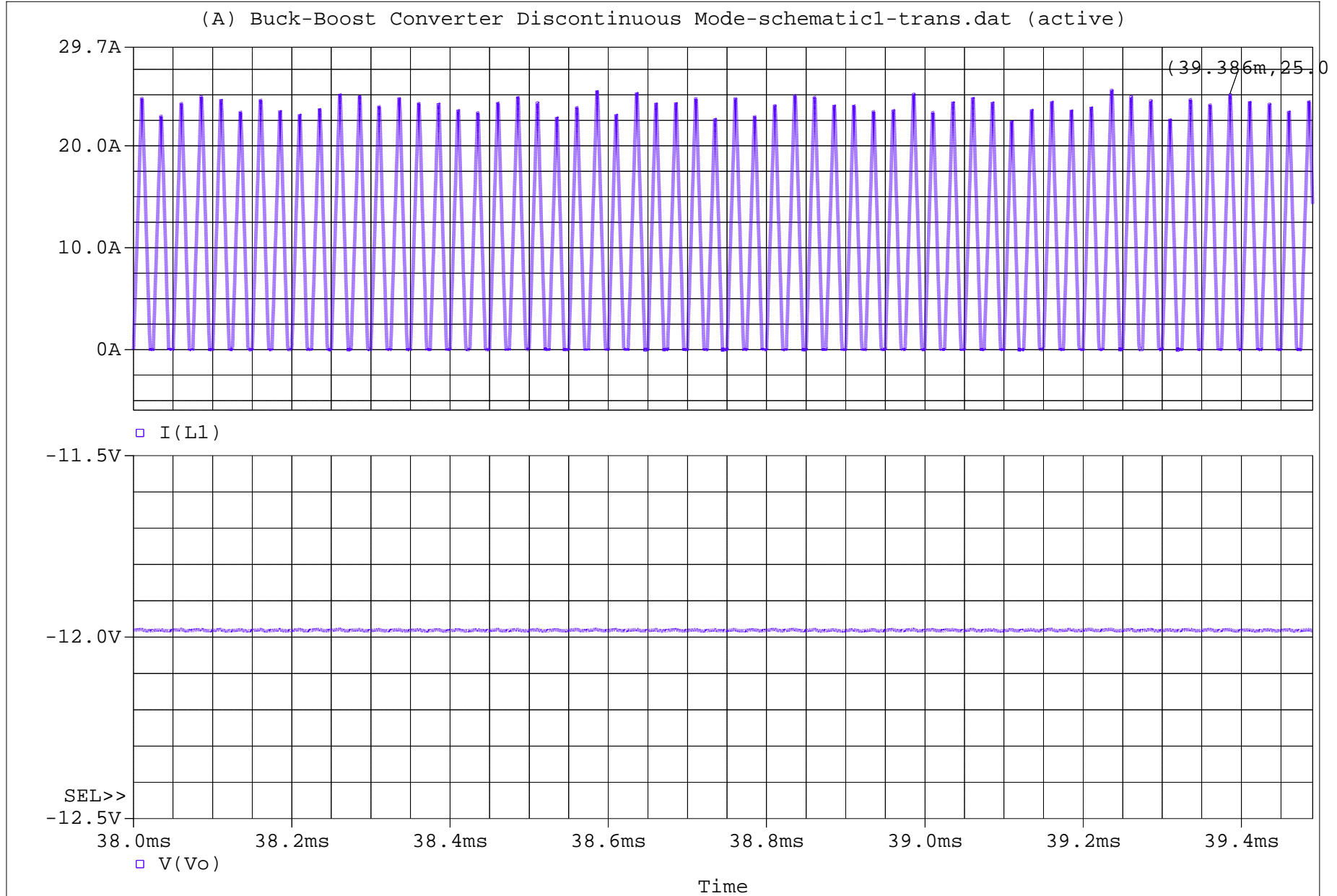
$$I_{\text{rms}} = 7.432 \text{ amp}$$

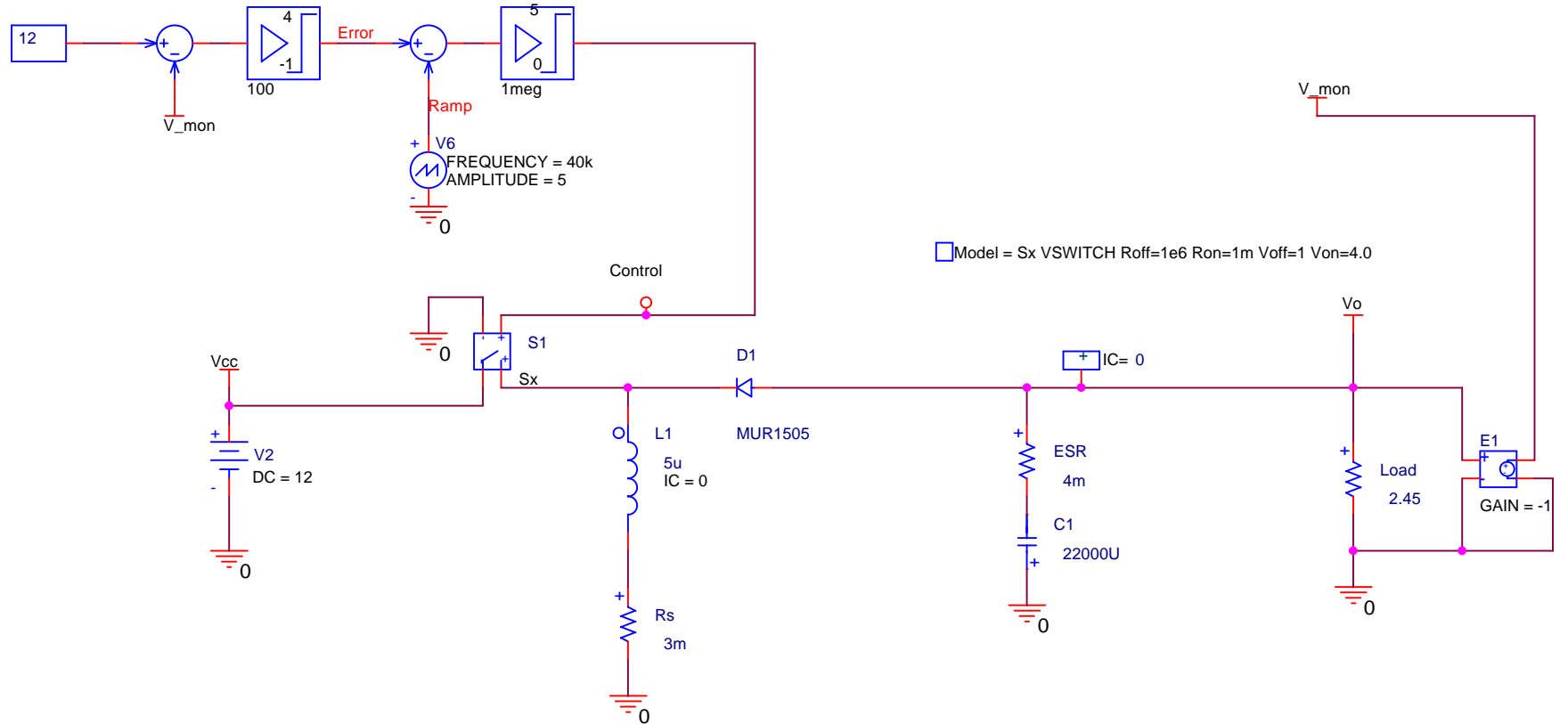



		ECE Department	
		5500 Wabash Avenue Terre Haute, IN 47803 Ph: (812) 877-8512 FAX: (253) 369-9536	
Name: Marc E. Herniter		Class: ECE456	
Size A	Document Name		Rev 1
Date: Monday, January 13, 2003		Sheet 1 of 1	

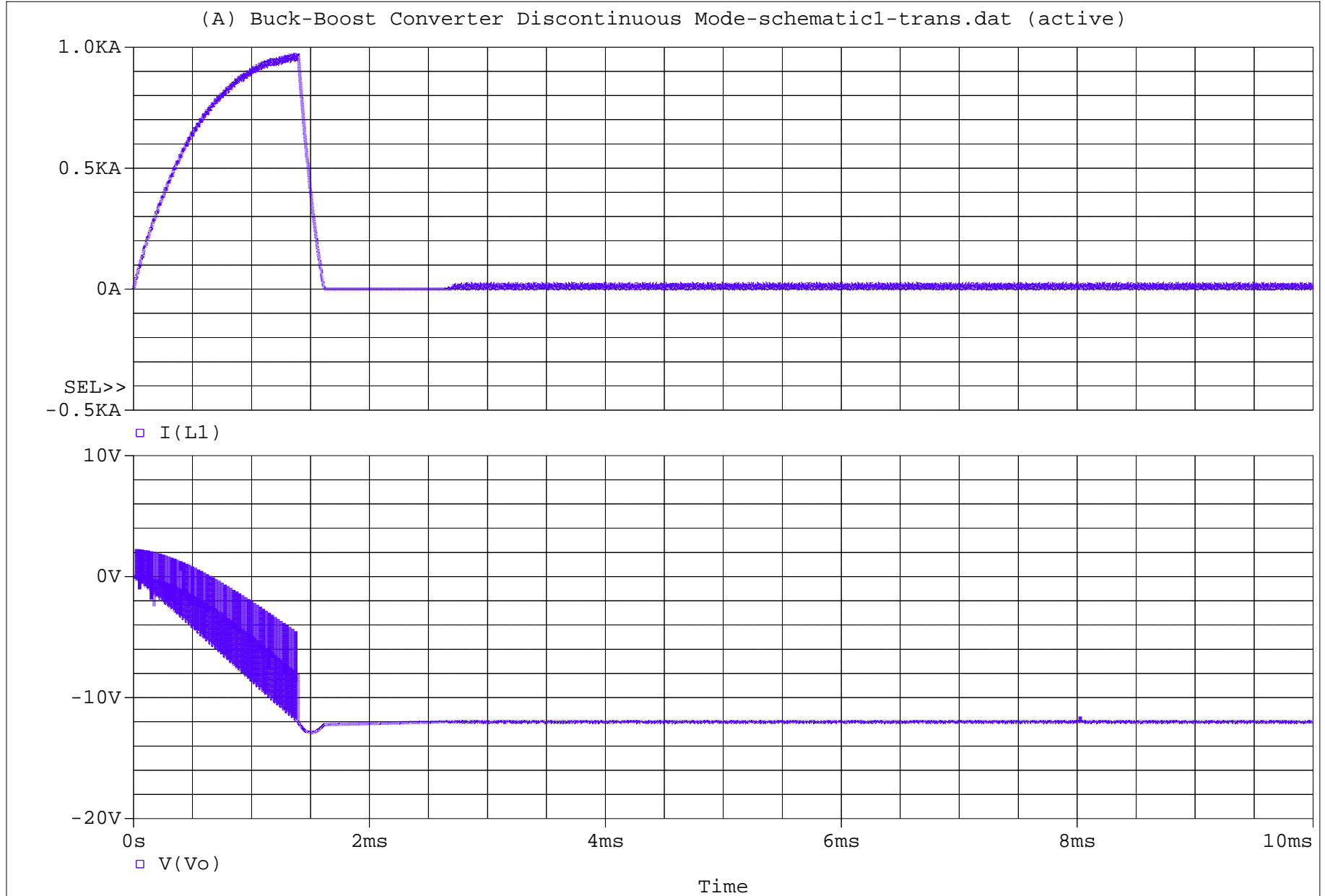








		ECE Department	
		5500 Wabash Avenue Terre Haute, IN 47803 Ph: (812) 877-8512 FAX: (253) 369-9536	
Name: Marc E. Herniter		Class: ECE456	
Size A	Document Name		Rev 1
Date: Monday, January 13, 2003		Sheet 1 of 1	



(A) Buck-Boost Converter Discontinuous Mode-schematic1-trans.dat (active)

